
Root Locus in the z-Domain

ECE 461/661 Controls Systems

Jake Glower - Lecture #31

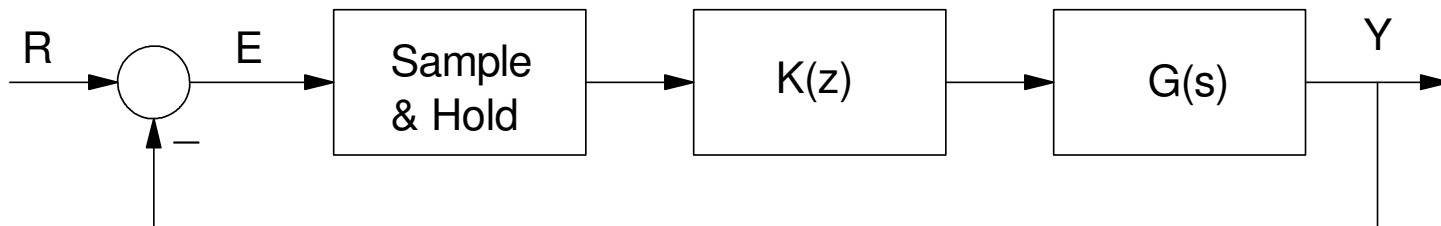
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lecture notes, homework sets, and solutions

Root Locus in the z-Domain

Goal: Find $K(z)$ for a "good" response

Note: Relative to the microcontroller

- The feedback system looks like it's a discrete-time system.
- It looks like it's in the z-domain.



Mathematically, the open-loop transfer function is

$$H(s) G(s) K(z)$$

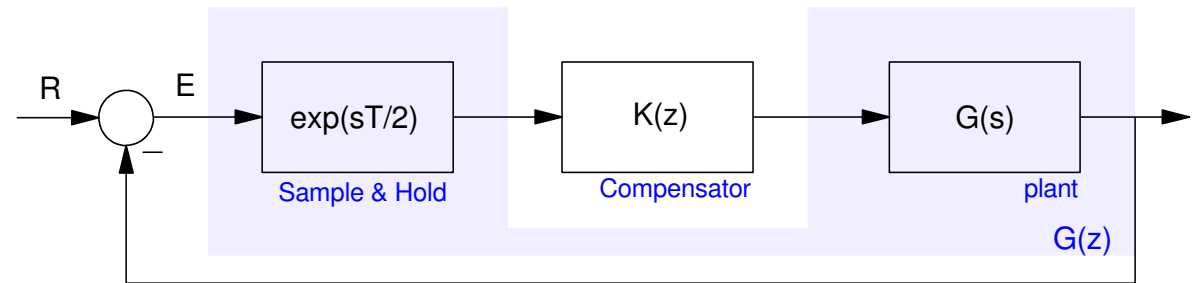
- $H(s)$ Sample and Hold
- $G(s)$ Analog Plant
- $K(z)$ Digital Controller

$H(s)$ looks like a 1/2 sample delay

$$H(s) \approx \exp\left(-\frac{sT}{2}\right)$$

Lump $H(s)$ and $G(s)$ together

$$H(s) G(s) \Rightarrow G(z)$$



The closed-loop system is

$$Y = \left(\frac{GK}{1+GK} \right) R$$

If GK has zeros and poles:

$$GK = k \frac{z}{p}$$

the closed-loop transfer function becomes

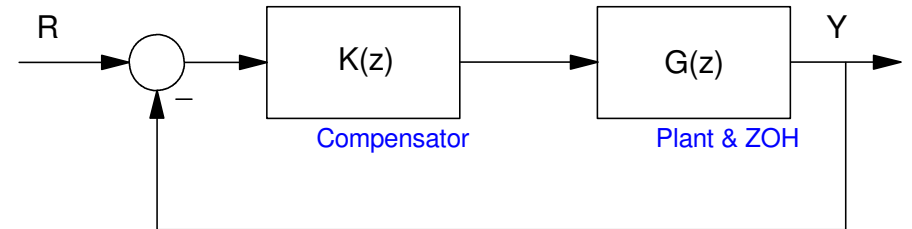
$$Y = \left(\frac{kz}{p+kz} \right) E$$

The roots of the closed-loop system are:

$$p(z) + k z(z) = 0$$

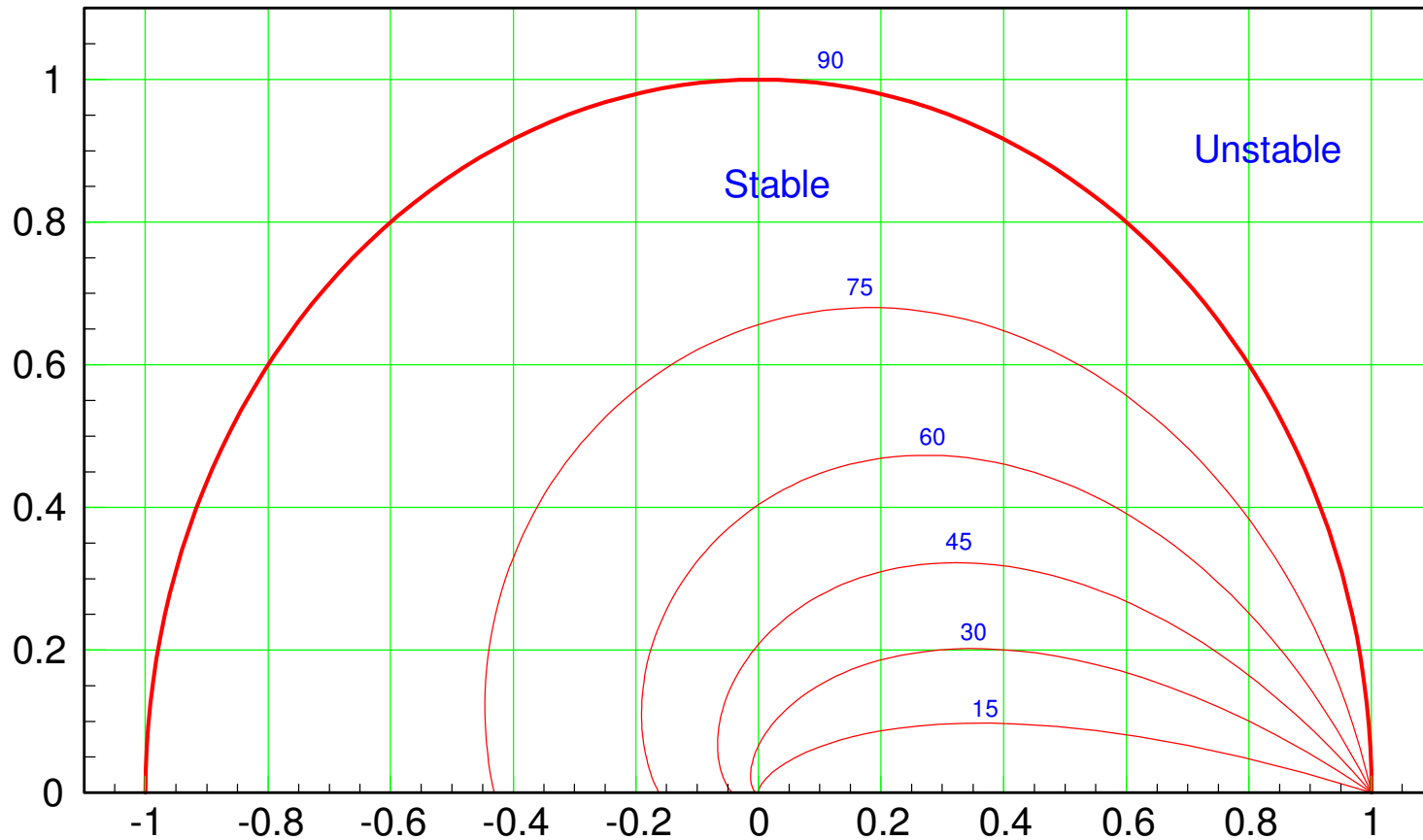
Note: This is identical to the s-plane

$$p(s) + k z(s) = 0$$



Root Locus in the z-plane

- Exactly the same as the s-plane
- The only difference is how to interpret the results: $z = e^{sT}$



2% Settling Time:

- Determined by the magnitude of the poles

$$z^k = 0.02$$

$$k = \frac{\ln(0.02)}{\ln(z)}$$

A pole at $z = 0.95$ has a 2% settling time of 73 samples (round up)

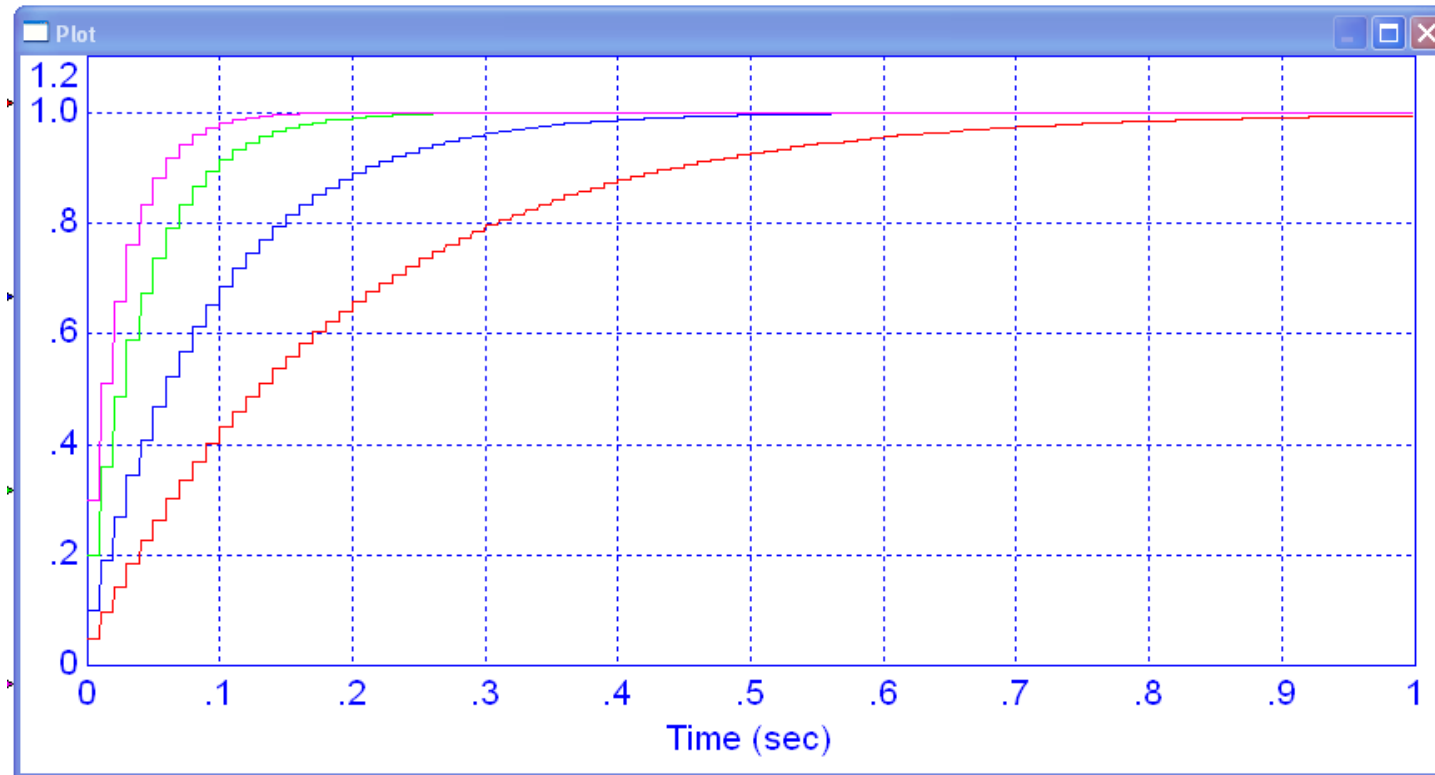
$$k = \frac{\ln(0.02)}{\ln(0.95)} = 72.26 \text{ samples}$$

A pole at $z = 0.8$ has a 2% settling time of 18 samples

$$k = \frac{\ln(0.02)}{\ln(0.8)} = 17.53 \text{ samples}$$

Similarly, for any pole, the amplitude tells you the settling time as

$$|z|^k = 0.02 \quad k = \frac{\ln(0.02)}{\ln(|z|)}$$



Step Response: Pole at { 0.95 (red), 0.9 (blue), 0.8 (green), and 0.7 (pink). T = 0.01 second

Frequency of Oscillation:

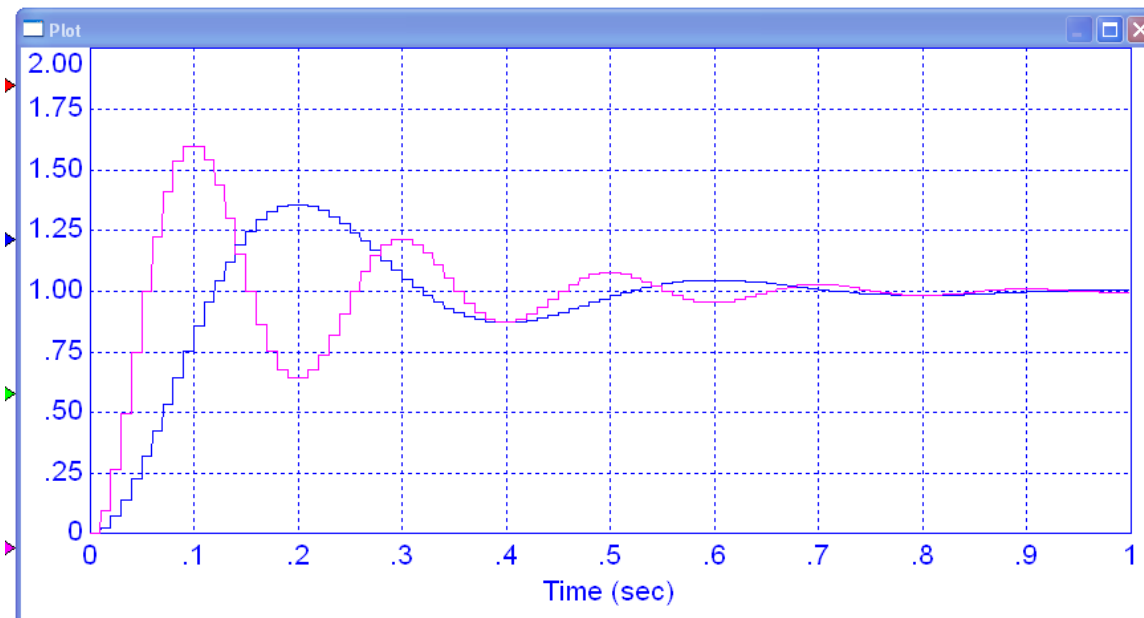
- Determined by the angle of the poles

$$\text{period} = \frac{360^\circ}{\text{angle}} \text{ samples}$$

Poles at

$$0.95 \angle \pm 9^\circ \quad \text{Period} = 40 \text{ samples}$$

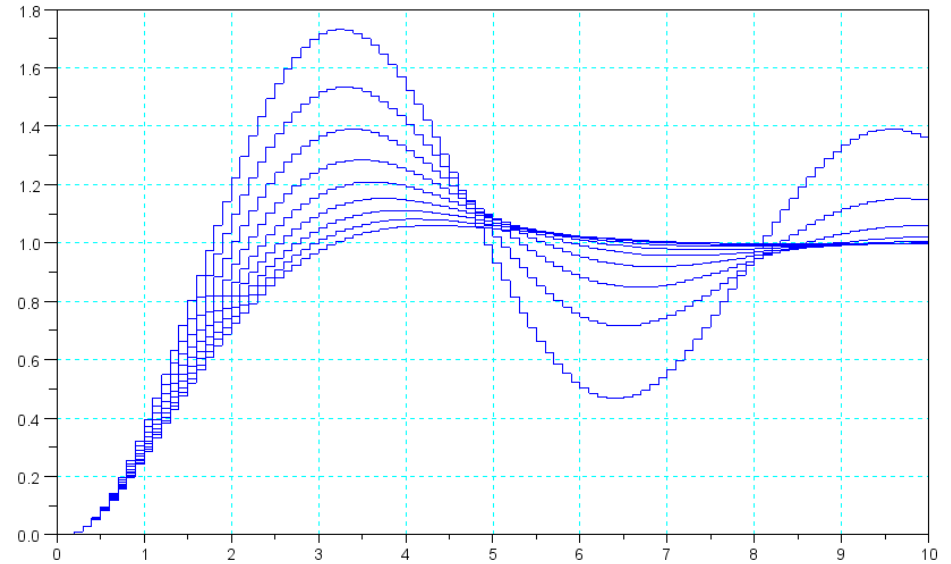
$$0.95 \angle \pm 18^\circ \quad \text{Period} = 20 \text{ samples}$$



Overshoot (Damping Ratio)

- Follows a log spiral
- Convert to the s-plane to find the damping ratio

Damping Ratio	s-plane	z-plane
0	$-1 \angle \pm 90^\circ$	$0.9950 + j0.0998$
0.2	$-1 \angle \pm 78.46^\circ$	$0.9755 + j0.0959$
0.4	$-1 \angle \pm 66.42^\circ$	$0.9568 + j0.0879$
0.6	$-1 \angle \pm 53.13^\circ$	$0.9388 + j0.0753$
0.8	$-1 \angle \pm 36.86^\circ$	$0.9214 + j0.0553$
1	$-1 \angle \pm 0^\circ$	$0.9048 + j0$



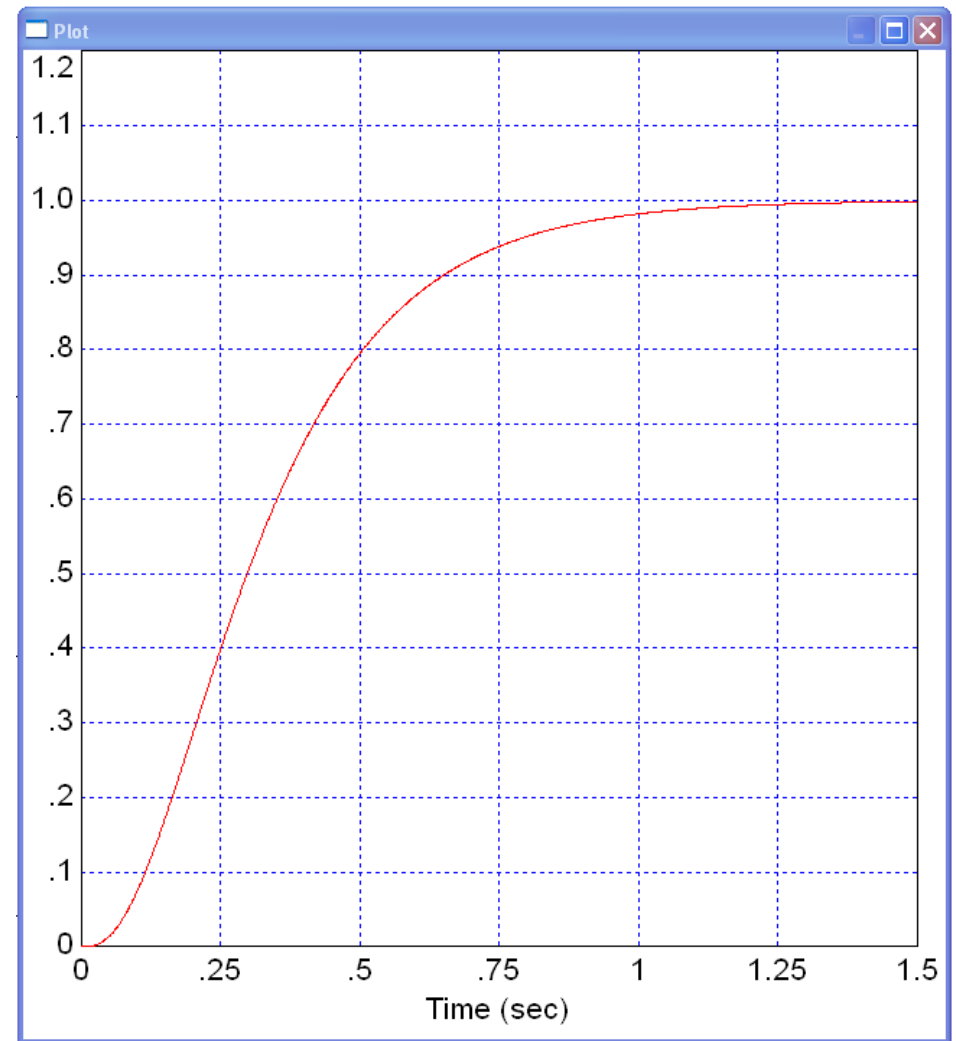
Gain Compensation in the z-Plane

Assume

$$G(s) = \left(\frac{1000}{(s+5)(s+10)(s+20)} \right)$$

Find $K(z) = k$

- $T = 10\text{ms}$
- No overshoot,
- 20% overshoot, and
- The maximum gain for stability.

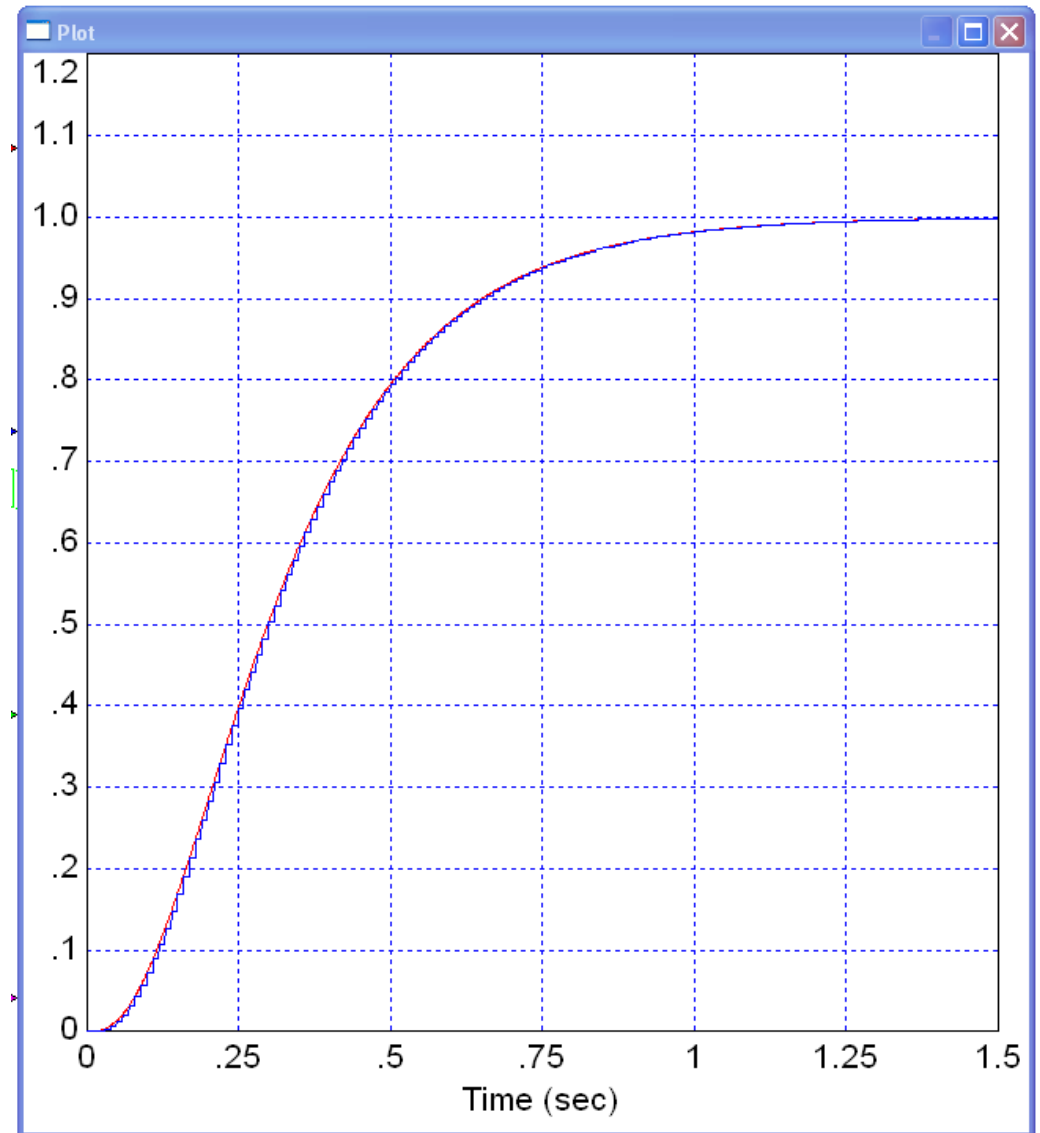


Step 1: Convert $G(s)$ to $G(z)$.

$$G(s) = \left(\frac{1000}{(s+5)(s+10)(s+20)} \right)$$

With $T = 0.01$ second

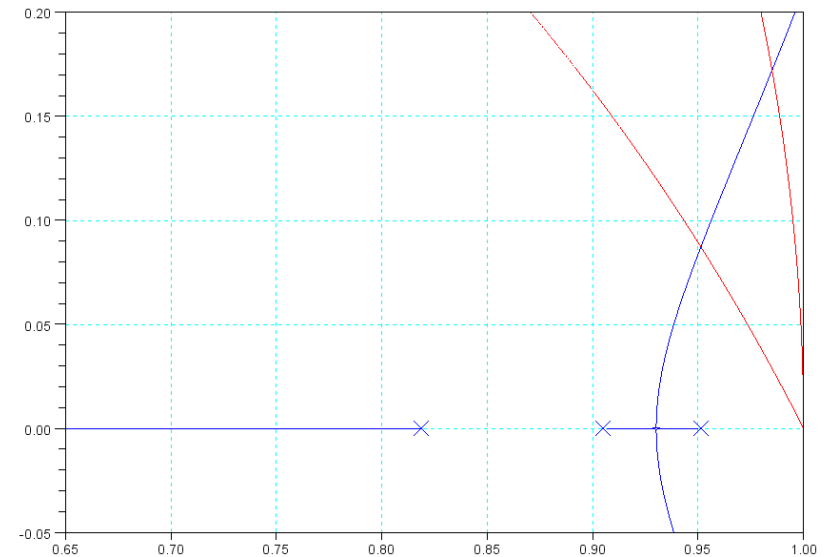
$$G(z) \approx \left(\frac{0.0008413z}{(z-0.9512)(z-0.9048)(z-0.8187)} \right)$$



Step 2: Draw the root locus of $G(z)$

- Add the damping line

```
T = 0.01;  
Gz = zpke(0, [0.9512, 0.9048, 0.8187],  
0.0008413);  
k = logspace(-2, 2, 1000)';  
R = rlocus(G, k);  
  
% draw the damping lines on this graph  
  
hold on  
s = [0:0.01:100] * (-1+j*2);  
z = exp(s*T);  
plot(real(z), imag(z), 'r')  
  
s = [0:0.01:100] * (j*1);  
z = exp(s*T);  
plot(real(z), imag(z), 'r')
```



Step 3: Pick a spot on the root locus

a) No overshoot.

- This is the breakaway point

$$z = 0.9305$$

At this point

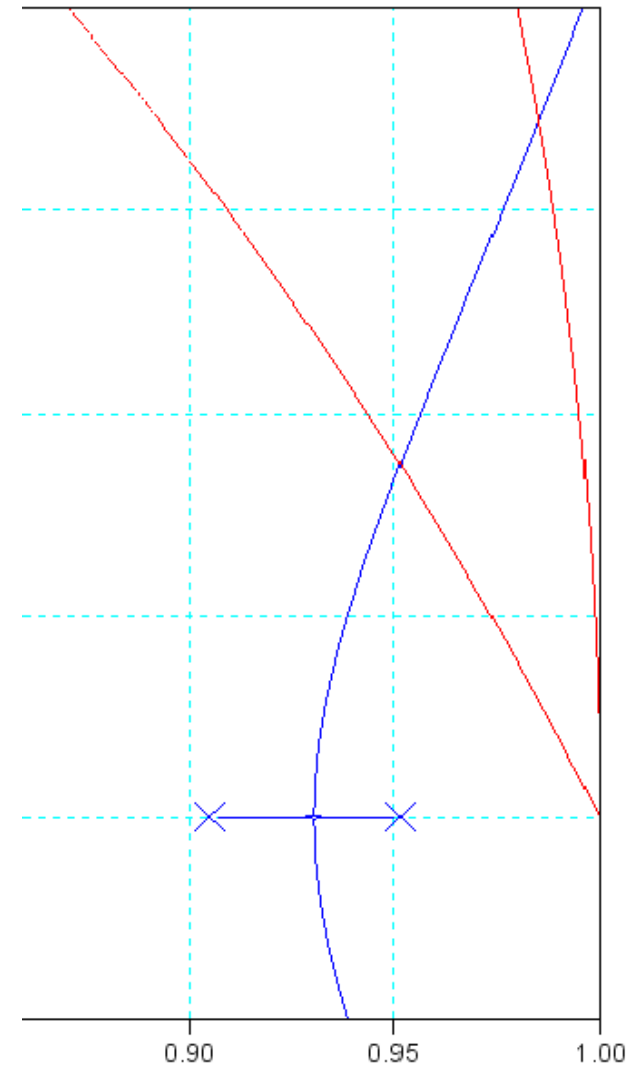
$$\left(\frac{0.0008413z}{(z-0.9512)(z-0.9048)(z-0.8187)} \right)_{z=0.9305} = 13.1620 \angle 180^\circ$$

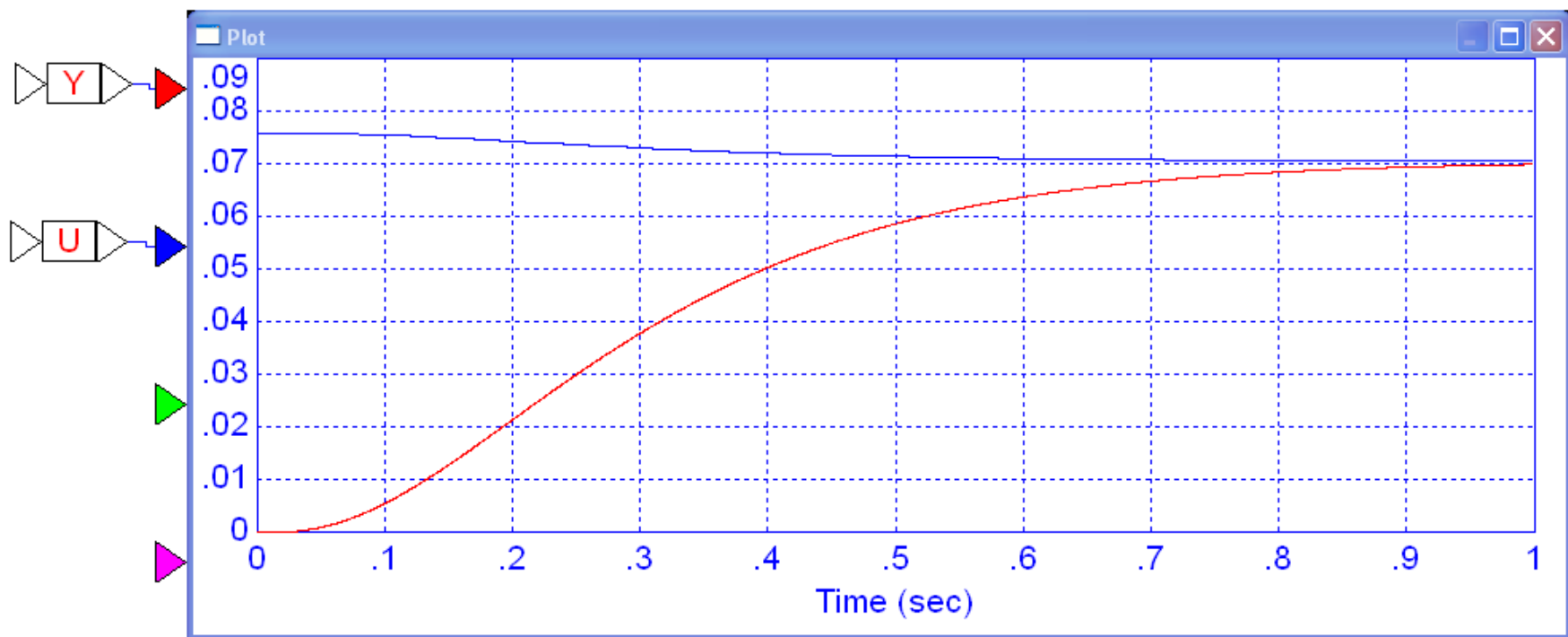
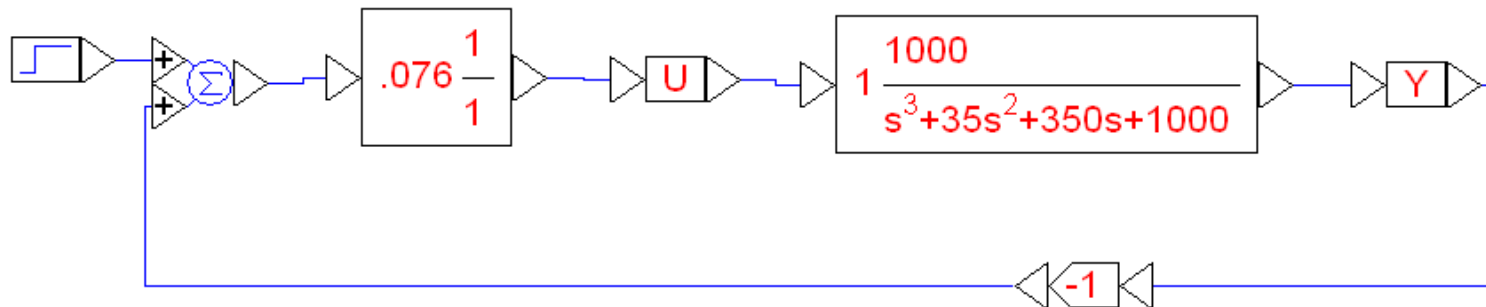
$$K = \frac{1}{13.1620} = 0.0760$$

This results in

$$t_{2\%} = \frac{\ln(0.02)}{\ln(0.9305)} = 54.3 \text{ samples (0.543 seconds)}$$

$$K_p = 0.0760$$





Step Response with $K(z)$ chosen to place the poles at the breakaway point (no overshoot).

b) 20% overshoot.

$$z = 0.9513 + j0.0873$$

At this point

$$\left(\frac{0.0008413z}{(z-0.9512)(z-0.9048)(z-0.8187)} \right)_{z=0.9513+j0.0873} = 0.5867 \angle 180^\circ$$

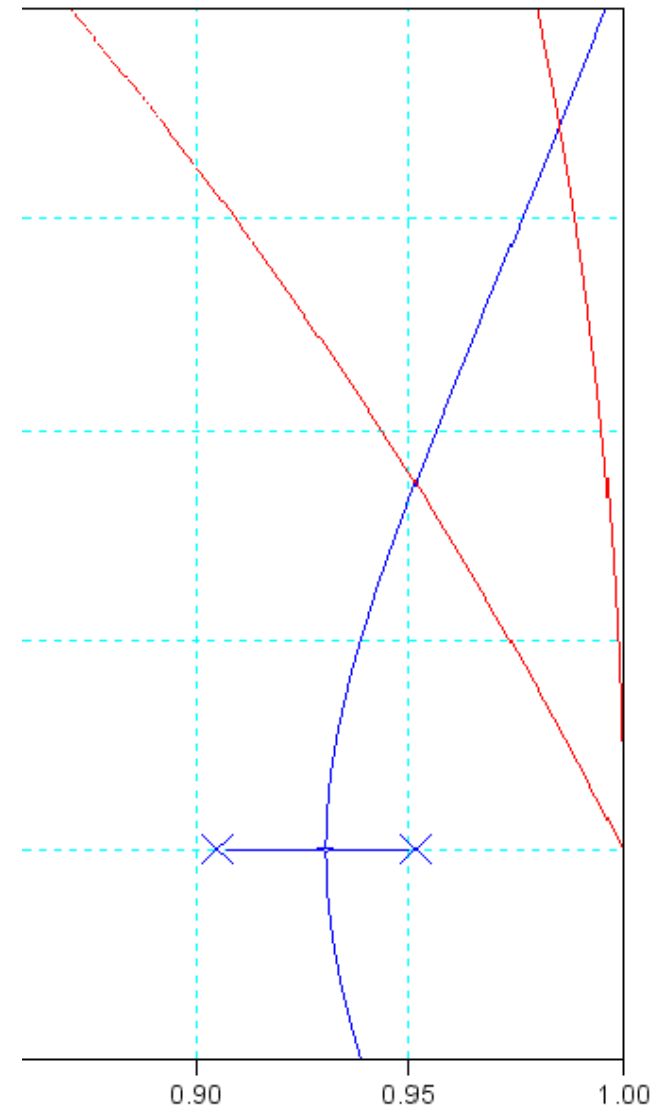
$K(z)$ is then

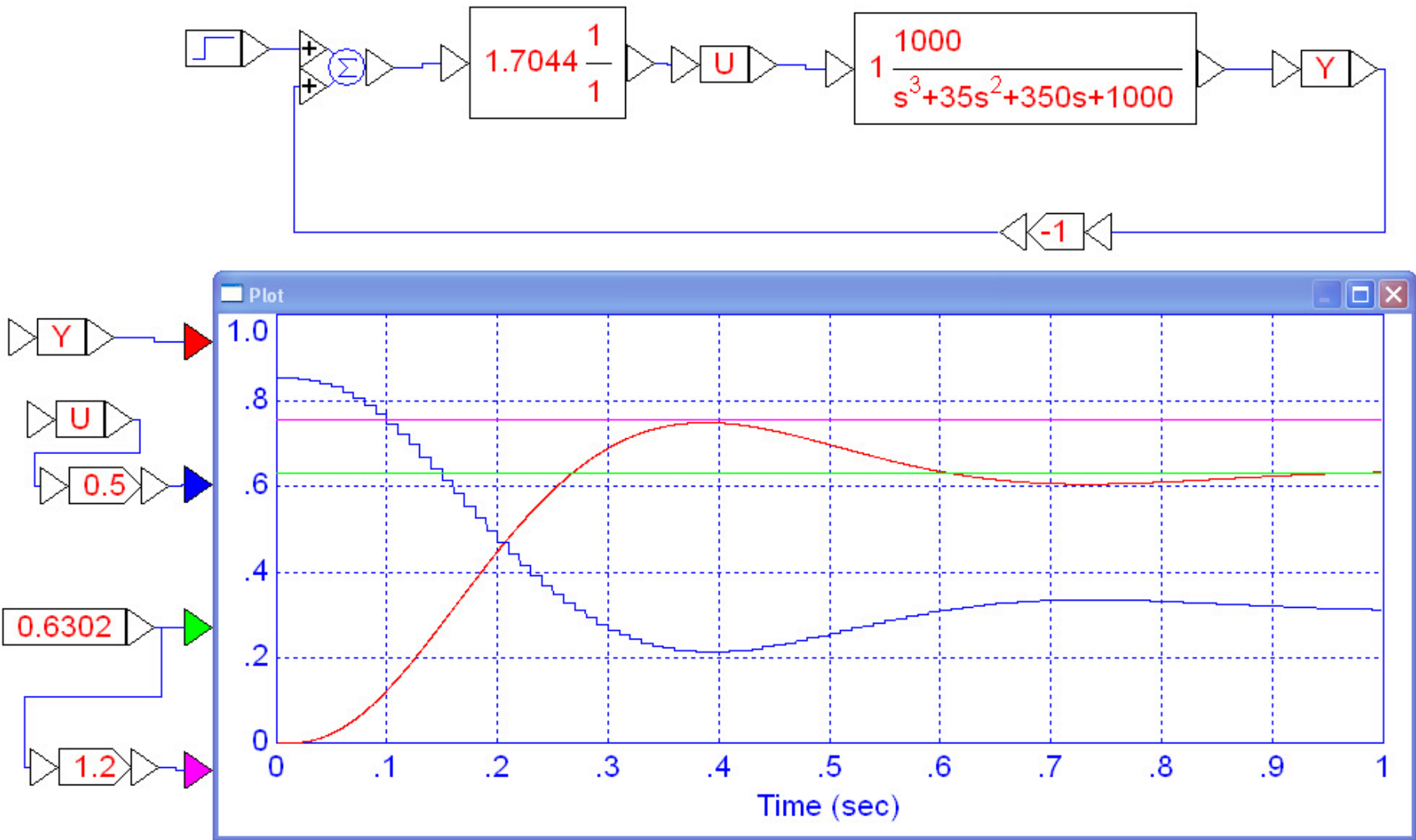
$$K = \frac{1}{0.5867} = 1.7044$$

This results in

$$t_{2\%} = \frac{\ln(0.02)}{\ln(|0.9513+j0.0873|)} = 85.52 \text{ samples}$$

$$K_p = 1.7044$$





Step Response with $K(z)$ chosen for 20% overshoot

c) Max Gain for Stability (the jw crossing).

$$z = 0.9850 + j0.1723$$

At this point

$$\left(\frac{0.0008413z}{(z-0.9512)(z-0.9048)(z-0.8187)} \right)_{z=0.9850+j0.1723} = 0.1053 \angle 180^\circ$$

meaning

$$K(z) = \frac{1}{0.1053} = 9.5008$$

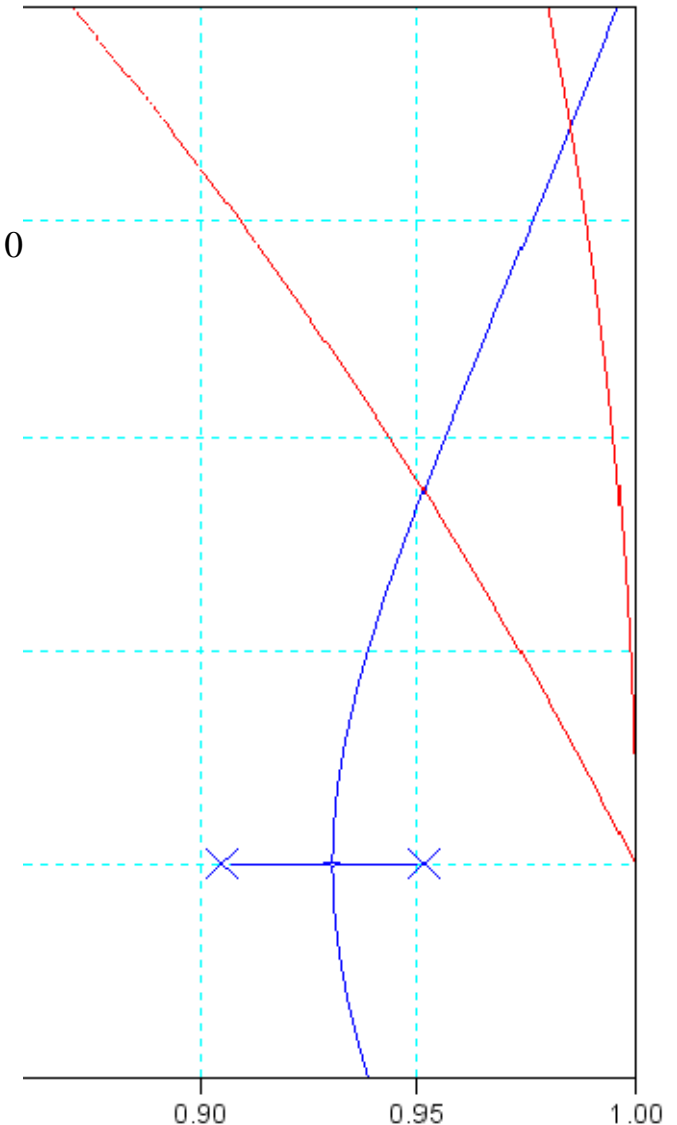
The frequency of oscillation is

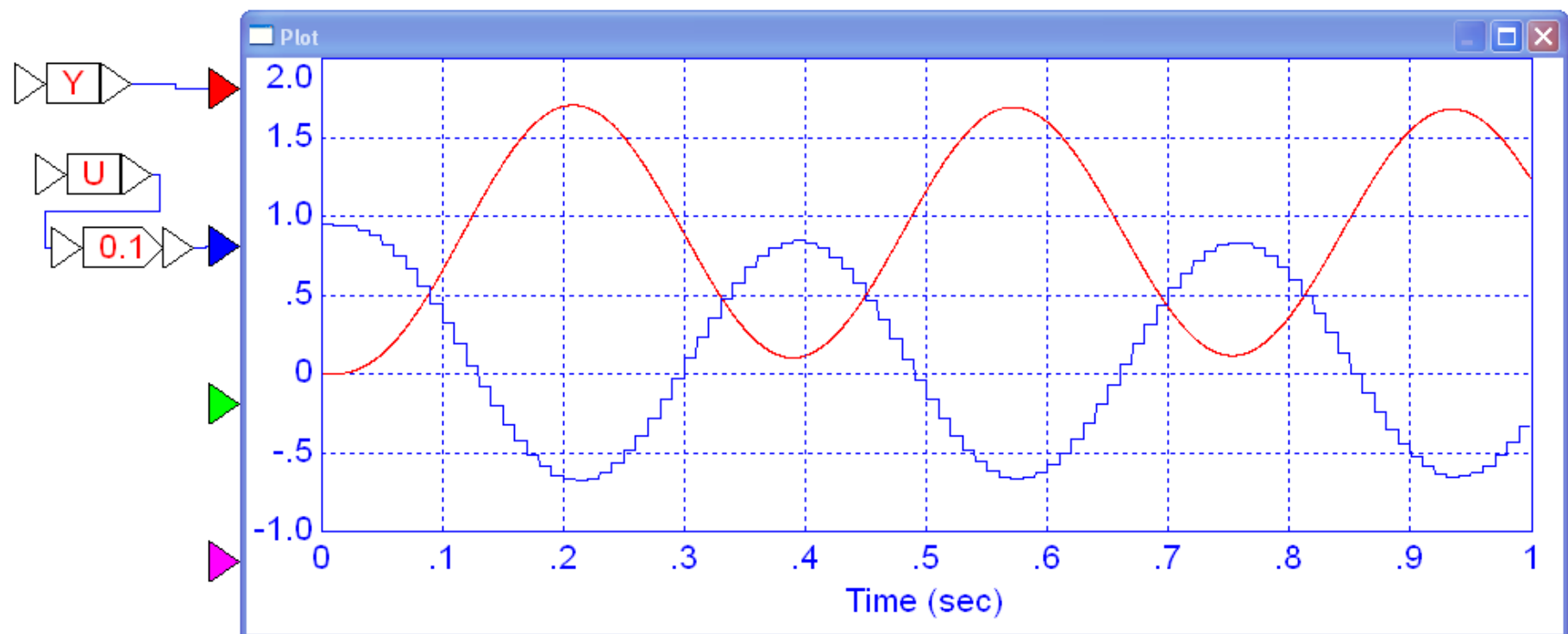
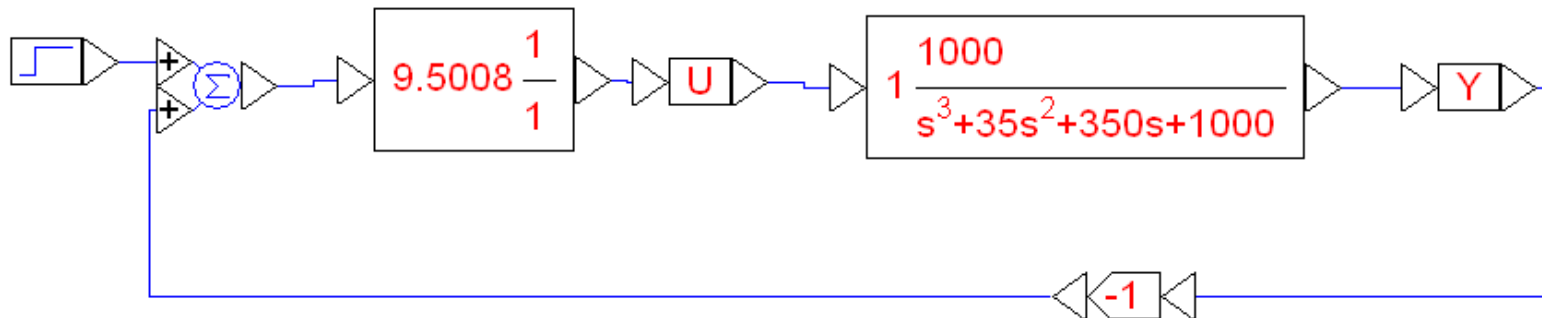
$$\angle z = 9.9215^\circ$$

$$\text{period} = \frac{360^\circ}{9.9215^\circ} = 36.28 \text{ samples}$$

$$= 0.3628 \text{ seconds}$$

$$f = \frac{1}{\text{period}} = 2.75 \text{ Hz}$$





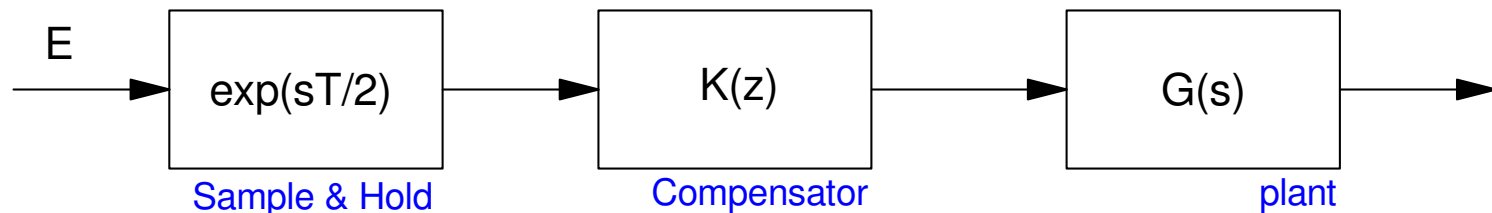
Alternate Method:

Find the spot on the damping line where the angles add up to 180 degrees

$$G \cdot K = 1 \angle 180^\circ$$

When you have a digital compensator, you really have three terms:

$$G(s) * K(z) * \text{sample and hold}$$



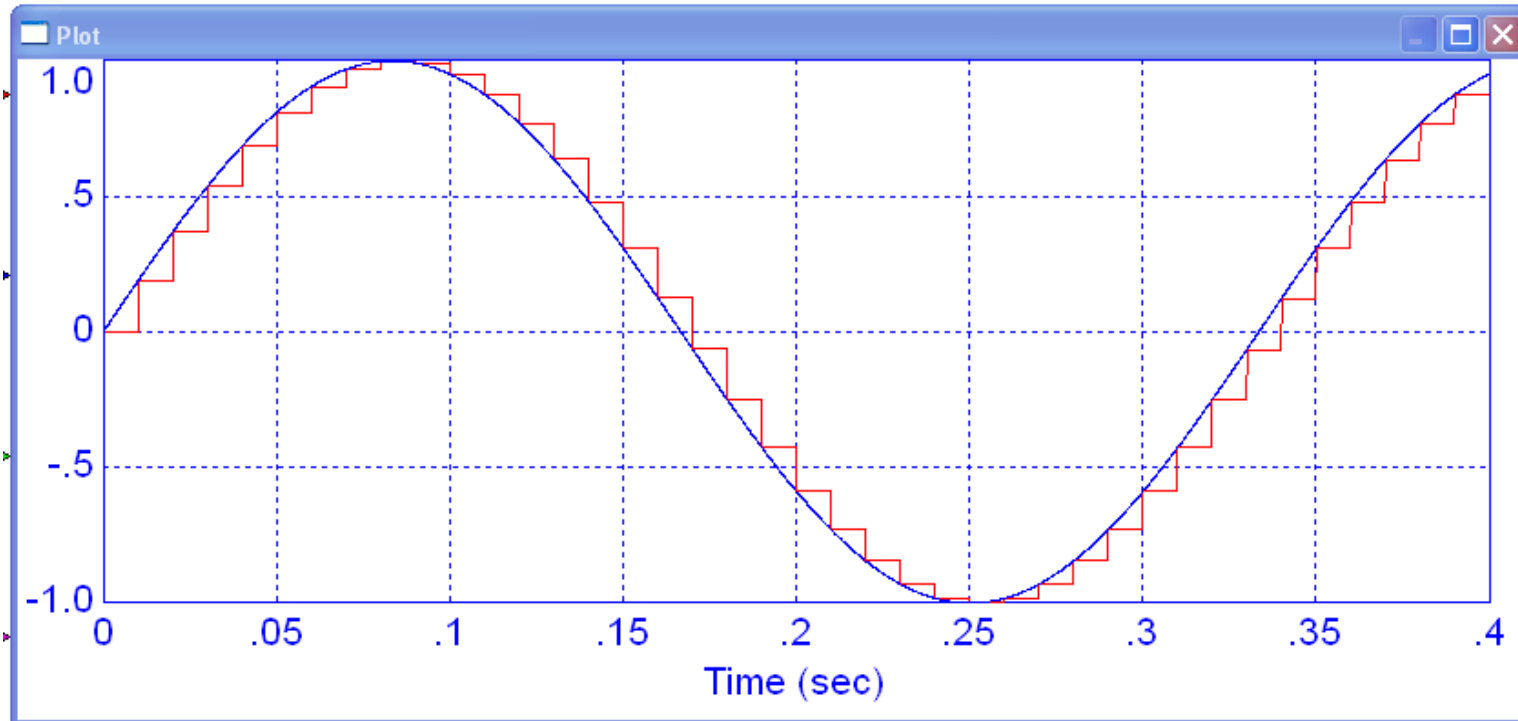
Matlab can analyze each of these

- You don't have to do s to z conversions
-

Modeling the Sample & Hold

- Sample and hold adds a 1/2 sample delay

$$H(s) \approx \exp\left(\frac{-sT}{2}\right)$$



3Hz sine wave (blue) sampled at 10ms (red) results in a 3Hz sine wave delayed by 1/2 of a sample (5ms)

Open-Loop System is then

$$G(s) \cdot K(z) \cdot \exp\left(\frac{-sT}{2}\right)$$

Find the point on the damping line where the phase is 180 degrees

$$\angle\left(G(s) \cdot K(z) \cdot \exp\left(\frac{-sT}{2}\right)\right) = 180^\circ$$

Note: You're kind-of mixing planes with this approach. Since you only care about one point, however, you don't care. You just

- Guess the point, s
 - Compute the corresponding point in the z -plane as $z = e^{sT}$
 - Evaluate the above function, and
 - Repeat until the angles add up to 180 degrees
-

Example: Find the gain, k, that results in 20% overshoot in the step response.

Guess #1: $s = -5 + j10$

```
T = 0.01;  
s = 5 * (-1 + j*2);  
z = exp(s*T);
```

```
1000 / ((s+5)*(s+10)*(s+20)) * exp(-s*T/2)  
  
- 0.5009862 + 0.0882455i
```

The angle isn't zero (the complex part is non-zero)

- Try a different s, such as 10% smaller

```
s = s*0.9;  
z = exp(s*T);  
1000 / ((s+5)*(s+10)*(s+20)) * exp(-s*T/2)  
  
- 0.5110062 + 0.0796359i
```

Keep going until the complex part is zero

- Angle is 180 degrees
- time passes.....

```
s = 0.9999*s;  
1000 / ( (s+5)*(s+10)*(s+20)) * exp(-s*T/2)
```

```
ans = -0.5853 - 0.0000i
```

Close enough. This gives

```
>> k = 1/abs(ans)
```

```
k = 1.7085
```

```
>> s
```

```
s = -4.5760 + 9.1519i
```

```
>> z
```

```
z = 0.9465 + 0.0950i
```

Note:

- The numerical method gives almost the same answer as root locus
- It's easier and is actually more accurate

no s to z conversions

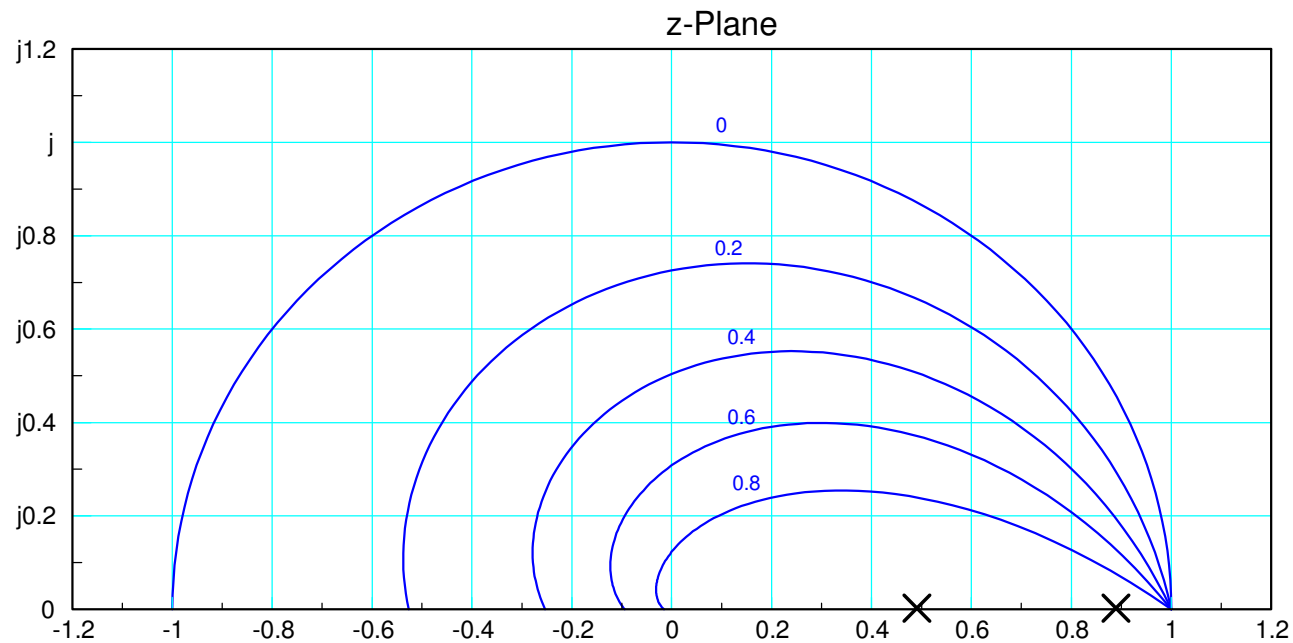
	Root Locus	Numerical Approach
k	1.7044	1.7085
s	-	-4.5760 + 9.1519i
z	0.9513 + j0.0873	0.9465 + 0.0950i

Handout: Sketch the root locus of

$$G(z) = \left(\frac{0.2}{(z-0.9)(z-0.5)} \right)$$

Find k for

- The breakaway point
- A damping ratio of 0.4
- The maximum gain for stability



Summary

- Root locus works in the s-plane
- Root locus works in the z-plane

The only difference is how you interpret the result:

$$T_s = \left(\frac{\ln(0.02)}{\ln(|z|)} \right)$$

$$period = \left(\frac{360^\circ}{\angle(z)} \right) \cdot T \text{ seconds}$$

$$z = e^{sT}$$
