
Lead Compensators using Root Locus

ECE 461/661 Controls Systems

Jake Glower - Lecture #23

Please visit [Bison Academy](#) for corresponding
lecture notes, homework sets, and solutions

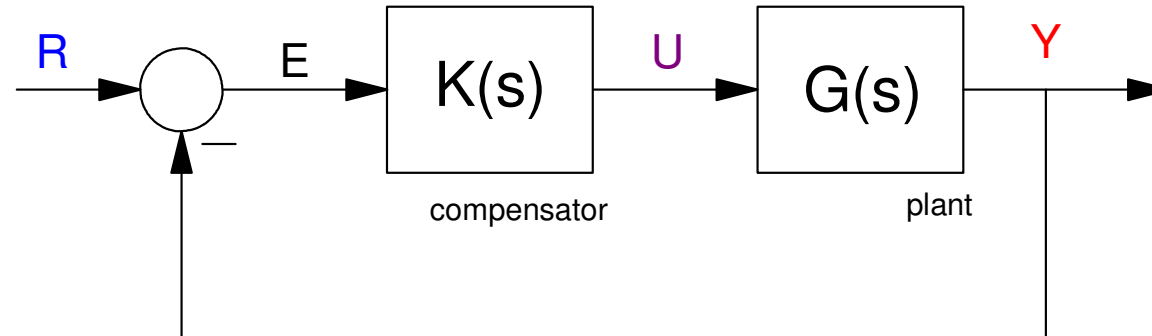
Introduction

Goal: Design a compensator, $K(s)$, to give

- Good tracking (error constants are proportional to k) and
- A fast response

Meaning...

- Make $K(s)$ large (faster system, better tracking),
- But not too large (stable and not too much overshoot).



Lead Compensator Design

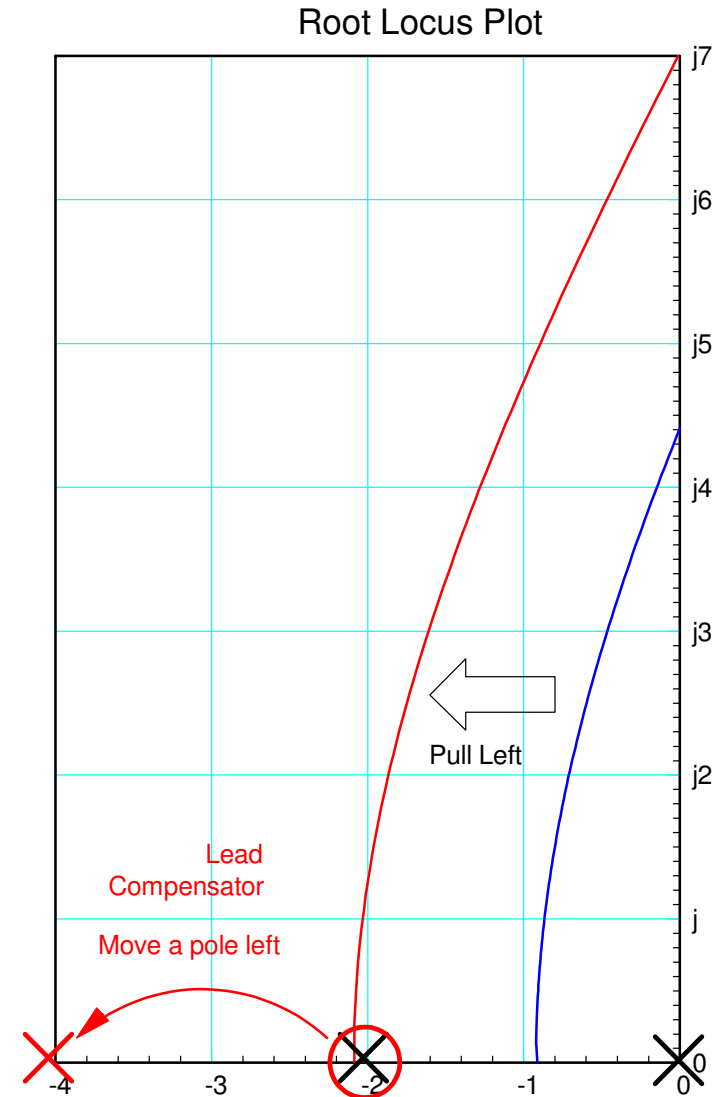
- $K(s) = k \left(\frac{s+a}{s+b} \right)$ $b > a$

Add a zero to cancel a pole

- Get rid of the one that's causing problems

Replace it with a faster pole

- Avoids differentiation (noise amplifier)
- Pulls the root locus left
- Speeds up the system



Which Pole do you Cancel?

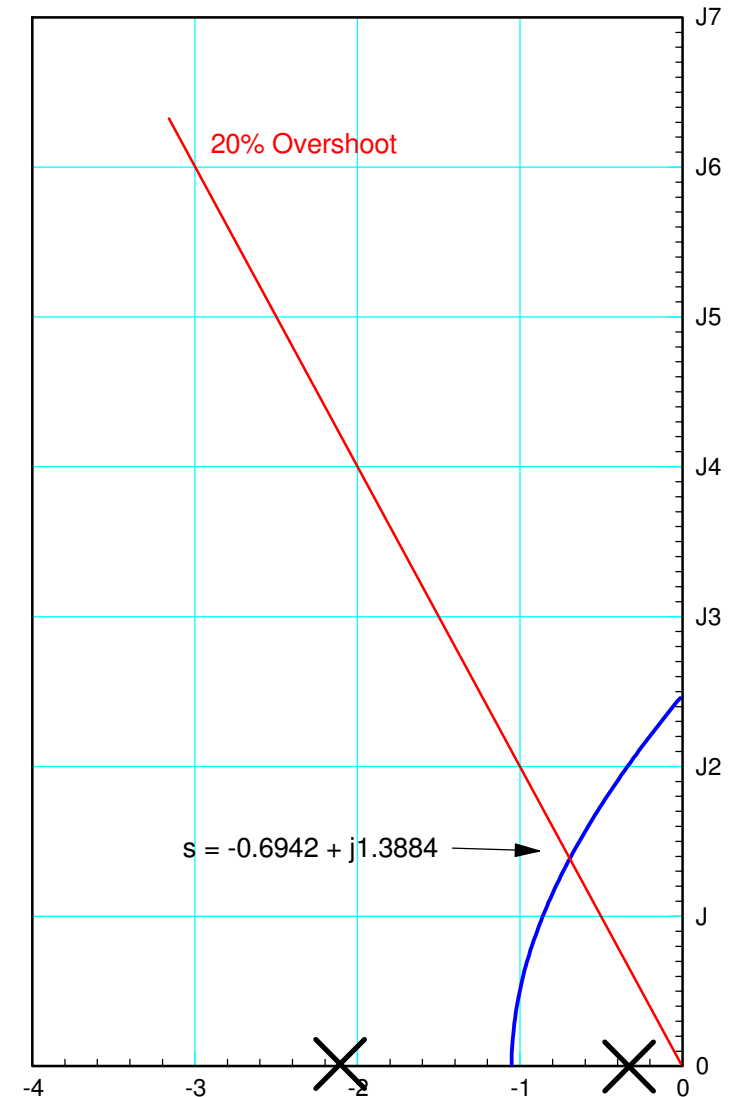
Example: 5th Order System

$$G(s) = \left(\frac{361.2378}{(s+0.3234)(s+2.081)(s+5.439)(s+10.1)(s+15.65)} \right)$$

Cancel Nothing (Gain Compensation)

For 20% Overshoot

- $s = -0.6942 + j1.3884$
- $K(s) = 5.5117$
- $K_p = 3.4412$
- $T_s = 5.76$ seconds



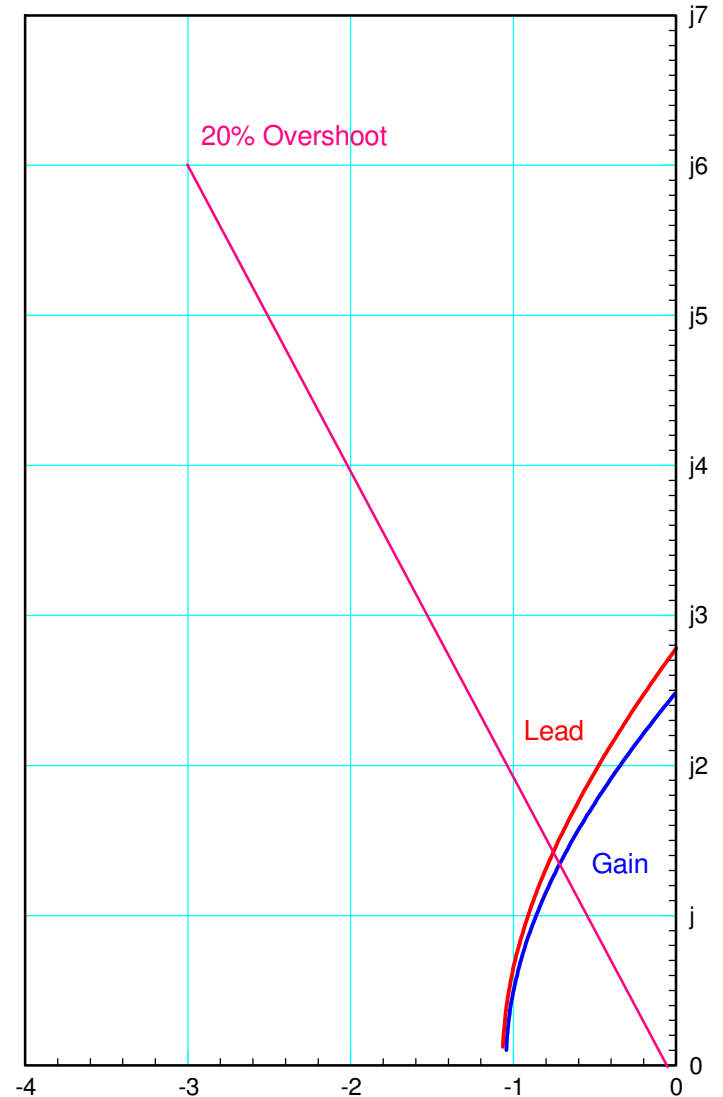
Cancel the Fastest Pole

$$G(s) = \left(\frac{361.2378}{(s+0.3234)(s+2.081)(s+5.439)(s+10.1)(s+15.65)} \right)$$

$$K(s) = k \left(\frac{s+15.65}{s+156.5} \right)$$

Result

- Almost no change
- The fast pole isn't the problem



Cancel slowest pole

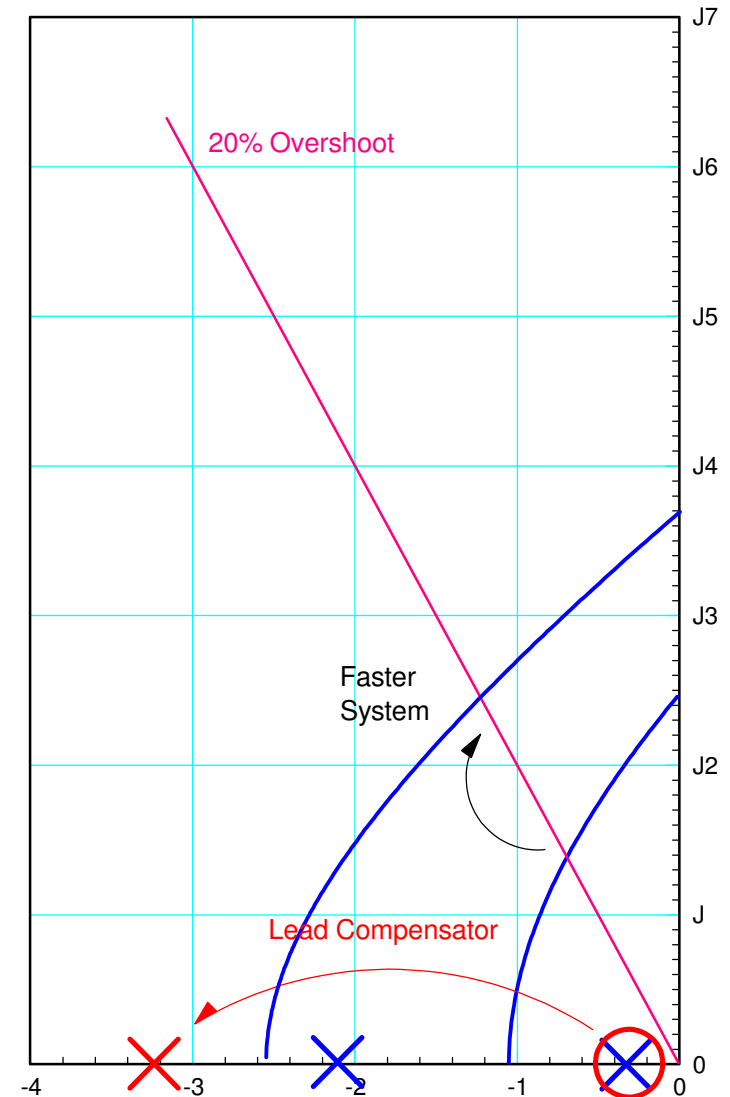
$$G(s) = \left(\frac{361.2378}{(s+0.3234)(s+2.081)(s+5.439)(s+10.1)(s+15.65)} \right)$$

$$K(s) = k \left(\frac{s+0.3242}{s+3.242} \right)$$

$$GK = \left(\frac{361.2378k}{(s+15.65)(s+10.1)(s+5.439)(s+2.081)(s+3.234)} \right)$$

For 20% Overshoot

- $s = -1.2531 + j2.5062$
- $K(s) = 15.30 \left(\frac{s+0.3242}{s+3.242} \right)$
- $K_p = 0.9554$ (worse)
- $T_s = 3.19$ sec (better)



Cancel the 2nd Slowest Pole

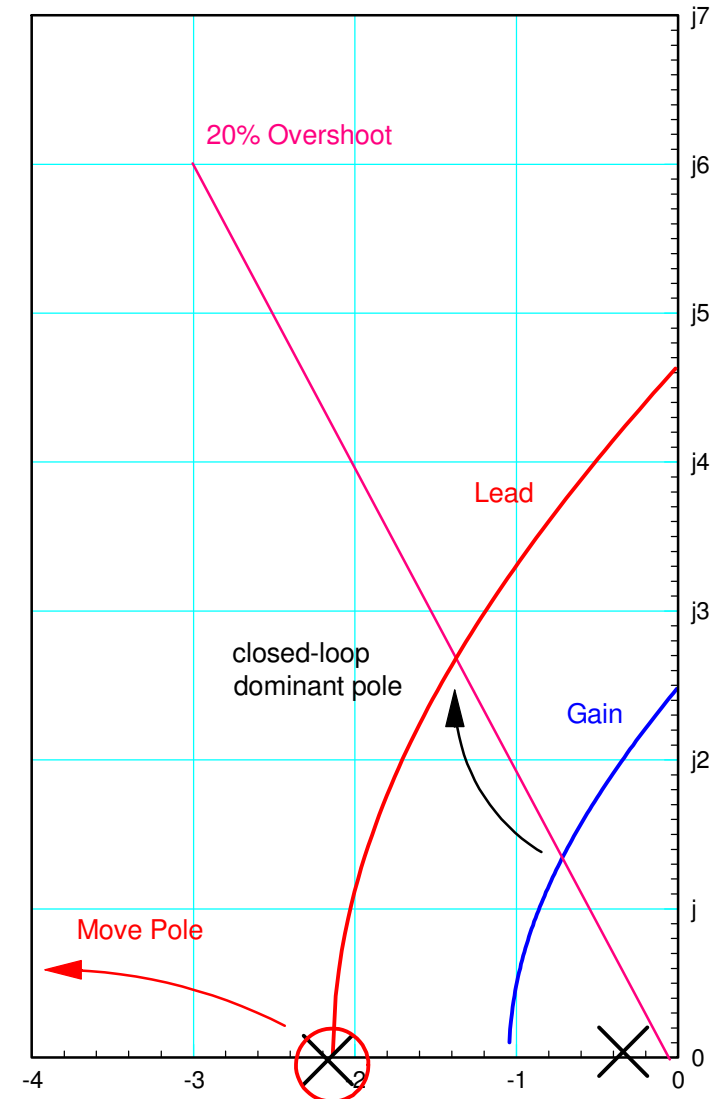
- Keep the pole at $s = -0.3234$
 - Like a pole at $s = 0$, reduces the steady-state error
- Cancel the pole at $s =$

$$G(s) = \left(\frac{361.2378}{(s+0.3234)(s+2.081)(s+5.439)(s+10.1)(s+15.65)} \right)$$

$$K(s) = k \left(\frac{2.081}{s+20.81} \right)$$

For 20% Overshoot

- $s = -1.3501 + j2.7002$
- $K(s) = 102.59 \left(\frac{s+2.081}{s+20.81} \right)$
- $K_p = 6.4052$ (2x better)
- $T_s = 2.9627$ (better)



Computations

$$K(s) = k \left(\frac{s+2.081}{s+20.81} \right)$$

$$GK = \left(\frac{361.2378k}{(s+20.81)(s+15.65)(s+10.1)(s+5.439)(s+0.3234)} \right)$$

$$\left(\frac{361.2378}{(s+20.81)(s+15.65)(s+10.1)(s+5.439)(s+0.3234)} \right)_{s=-1.3501+j2.7002} = 0.0097 \angle 180^\circ$$

$$k = \frac{1}{0.0097} = 102.59$$

$$K(s) = 102.59 \left(\frac{s+2.081}{s+20.81} \right)$$

$$T_{2\%} = \frac{4}{1.3501} = 2.9627 \text{ seconds}$$

$$K_p = (G(s) \cdot K(s))_{s=0} = 6.4052$$

Results:

Note that canceling the pole at -2.081 resulted in

- A slightly faster system
- With a much larger error constant, K_p , meaning better tracking
- As well as a much larger value of U at $t=0$

Lead Compensation $K(s) = k (s+a) / (s+10a)$				
$K(s)$	Closed-Loop Dominant Pole(s)	U at $t=0$ <small>$K(s)$ as $s \rightarrow \infty$</small>	K_p	2% Settling Time seconds
5.5117	$s = -0.6942 + j1.3884$	5.5117	3.4412	5.76
$15.30 \left(\frac{s+0.3242}{s+3.242} \right)$	$s = -1.2531 + j2.5062$	15.30	0.9554	3.19
$102.59 \left(\frac{s+2.081}{s+20.81} \right)$	$s = -1.3501 + j2.7002$	102.59	6.4052	2.96

Lead Compensator Design

Keep one pole at or near $s=0$.

- Pick the zero of the lead compensator to cancel the next slowest (stable) pole.
- Keeps K_p (or K_v) large

Pick the pole 3 to 10 times larger than the zero

- Speeds up the system
 - There is *some* limit on how much input you can apply
 - Assume 3x to 10x faster
-

Resulting Step Response

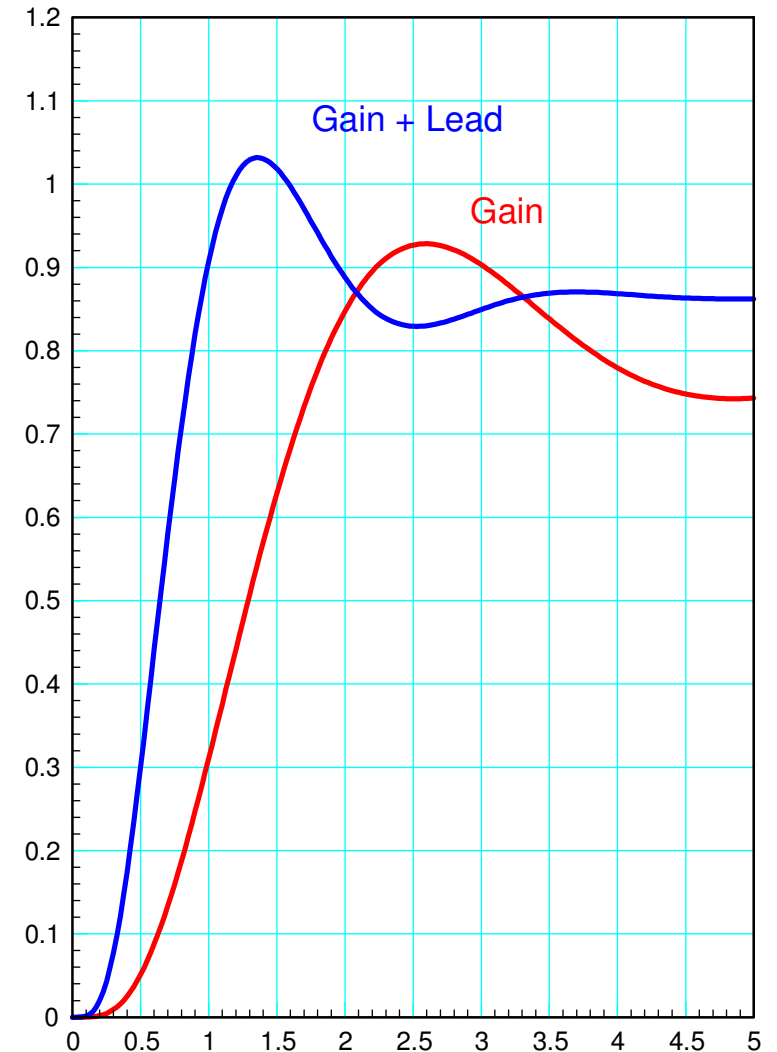
```
>> G = zpk([], [-0.3243, -2.081, -5.439, -10.1, -15.65], 361.2378);  
>> K = zpk(-2.081, -20.81, 101.61)
```

$$\frac{101.61 (s+2.081)}{(s+20.81)}$$

```
>> t = [0:0.01:5]';  
>> Gcl = minreal(G*K / (1+G*K));  
>> y = step(Gcl, t);  
>> plot(t, y);
```

Gain + Lead is

- Faster
 - root locus is further left
- With better tracking
 - Less steady-state error
 - Larger error constant



Lead Compensator Circuit: $K(s) = 101.61 \left(\frac{s+2.081}{s+20.81} \right)$

3 constraints, 4 degrees of freedom

- Assume $R_2 = 1\text{M}$

High frequency gain ($s \rightarrow \infty$)

$$K(s) = 101.61 = \frac{R_2}{R_a}$$

$$R_a = 9.84\text{k}$$

DC Gain ($s \rightarrow 0$)

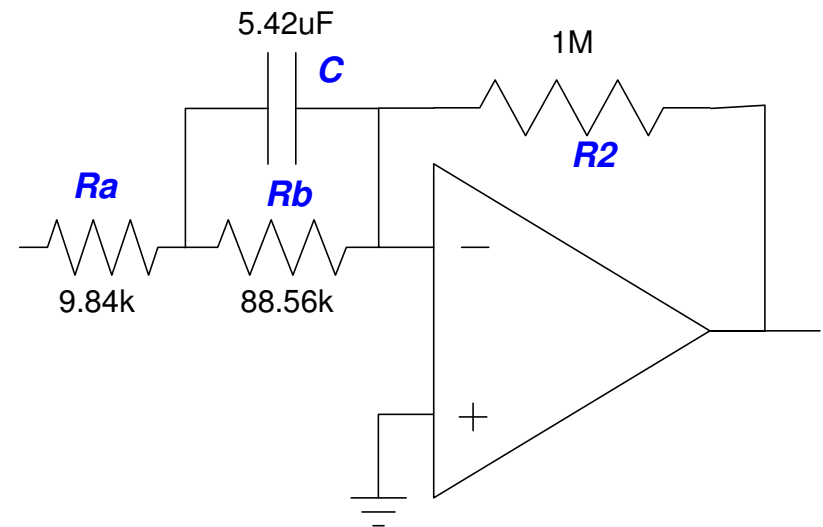
$$K(s) = 10.161 = \frac{R_2}{R_a + R_b}$$

$$R_b = 88.56\text{k}$$

Zero:

$$\frac{1}{R_b C} = 2.081$$

$$C = 5.42\mu\text{F}$$



Lead Compensator: Software

$$K(s) = 101.61 \left(\frac{s+2.081}{s+20.81} \right) = 101.61 \left(1 - \frac{18.729}{s+20.81} \right)$$

Add a dummy state, Z, which is

```
Z = 0;
```

```
R = 100;
```

```
while(t < 100)
```

```
    E = R - V(10);
```

```
    dZ = -20.81*Z + 18.729*E;
```

```
    V0 = 101.61 * ( E - Z );
```

```
    dV(1) = 10*V0 - 20.1*V(1) + 10*V(2);
```

```
    dV(2) = 10*V(1) - 20.1*V(2) + 10*V(3);
```

```
    dV(3) = 10*V(2) - 20.1*V(3) + 10*V(4);
```

```
    :
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```
    :
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Lead Compensators x 2

- If one lead compensator is good, why not two?

Keep the slowest pole ($s = -0.3234$)

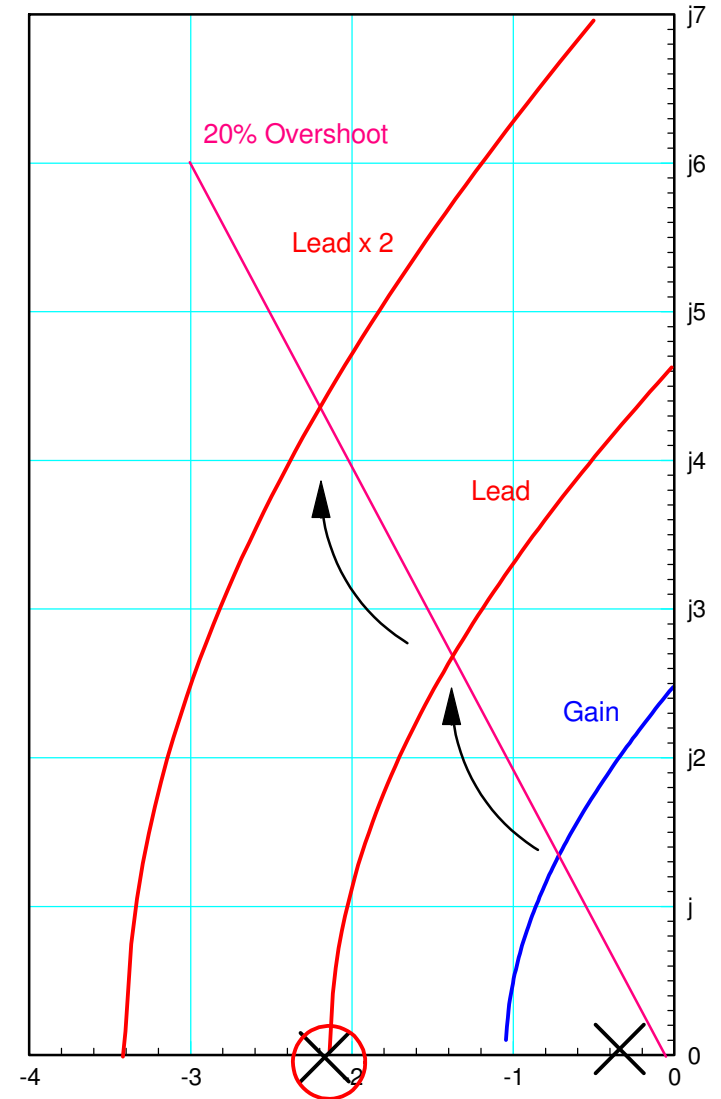
Cancel the next two slowest poles

$$K(s) = k \left(\frac{s+2.081}{s+20.81} \right) \left(\frac{s+5.439}{s+53.39} \right)$$

$$GK = \left(\frac{361.2378k}{(s+15.65)(s+10.1)(s+54.39)(s+20.81)(s+0.3234)} \right)$$

Result:

- $s = -2.2463 + j4.4925$
- $K(s) = 1729.60 \left(\frac{s+2.081}{s+20.81} \right) \left(\frac{s+5.439}{s+53.39} \right)$

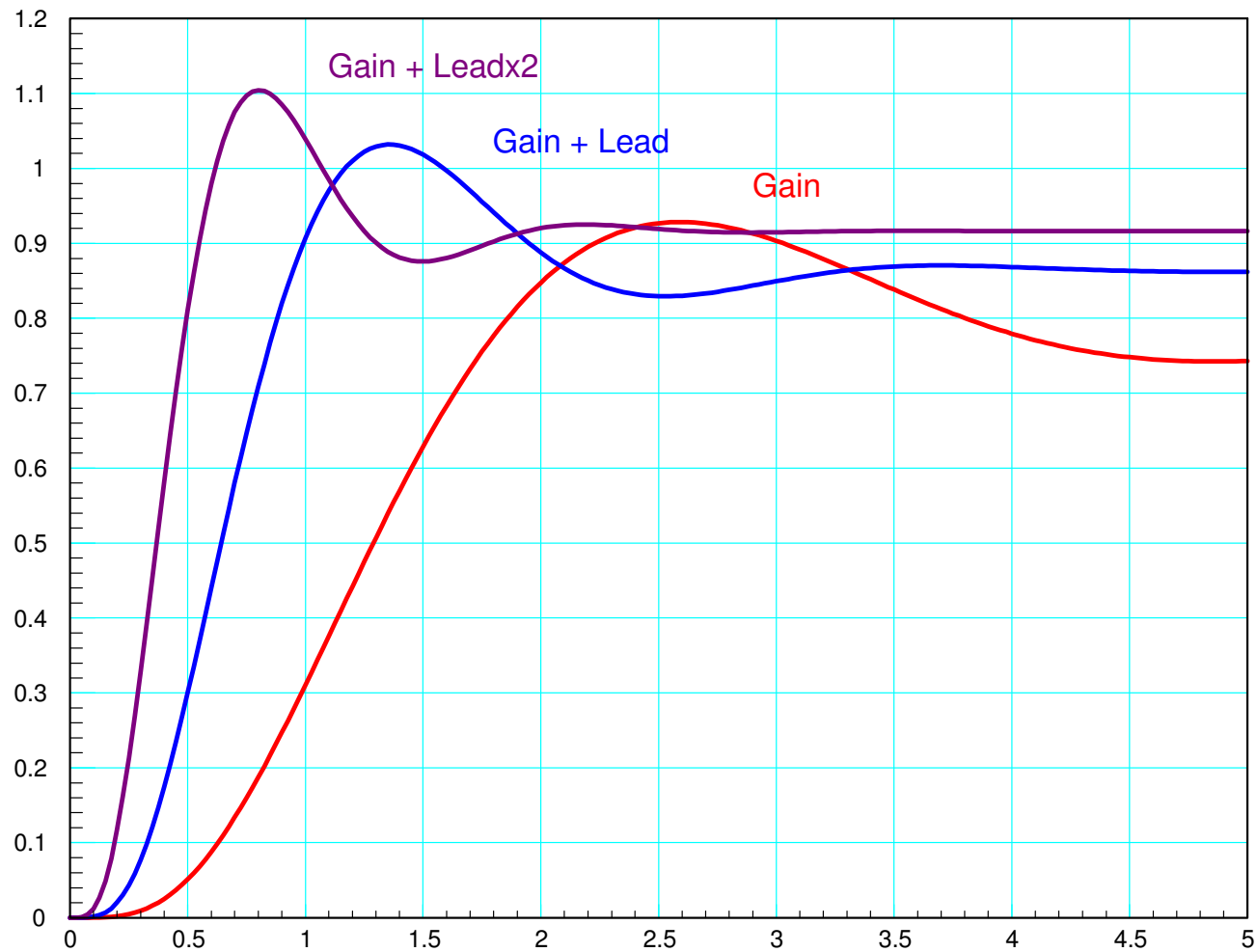


Result

- Each lead compensator speeds up the system
- The input at $t = 0$ increases 31x though...

Lead Compensation $K(s) = k (s+a) / (s+10a)$				
$K(s)$	Closed-Loop Dominant Pole(s)	U at $t = 0$ <small>$K(s)$ as $s \rightarrow \infty$</small>	K_p	$T_{2\%}$ seconds
5.5117	$s = -0.6942 + j1.3884$	5.5117	3.4412	5.76
$15.30 \left(\frac{s+0.3242}{s+3.242} \right)$	$s = -1.2531 + j2.5062$	15.3	0.9554	3.19
$102.59 \left(\frac{s+2.081}{s+20.81} \right)$	$s = -1.3501 + j2.7002$	102.59	6.4052	2.96
$1729.60 \left(\frac{s+2.081}{s+20.81} \right) \left(\frac{s+5.439}{s+53.39} \right)$	$s = -2.2463 + j4.4925$	1729.60	10.7987	1.78

Closed-Loop Response (Gain, Lead, Lead x 2):

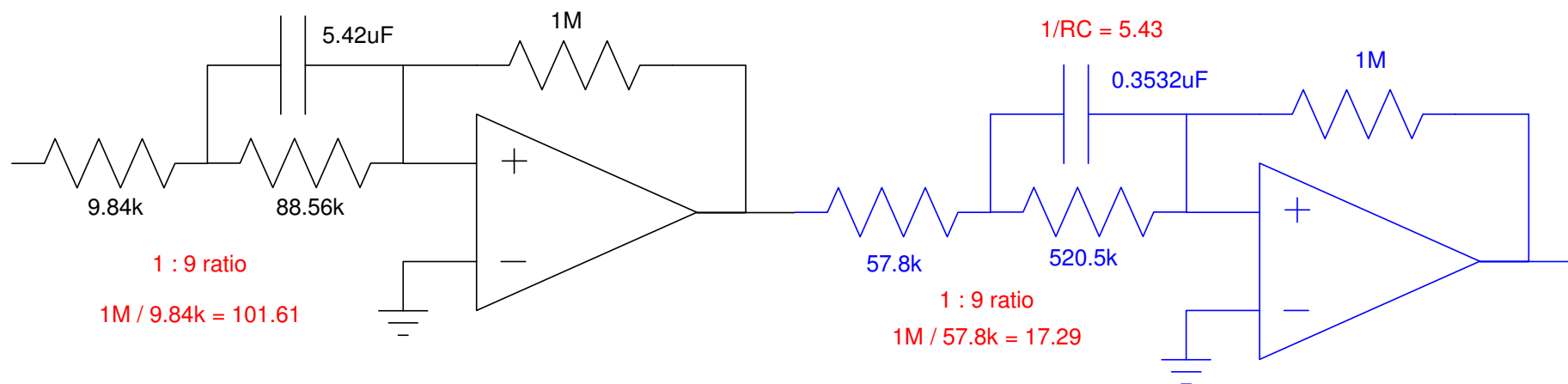


Step Response for the Gain, Lead, and Two Lead Compensated Systems

Lead Circuit

$$K(s) = 1729.60 \left(\frac{s+2.081}{s+20.81} \right) \left(\frac{s+5.439}{s+54.39} \right)$$

$$K(s) = \left(101.61 \frac{s+2.081}{s+20.81} \right) \left(17.29 \frac{s+5.439}{s+54.39} \right)$$

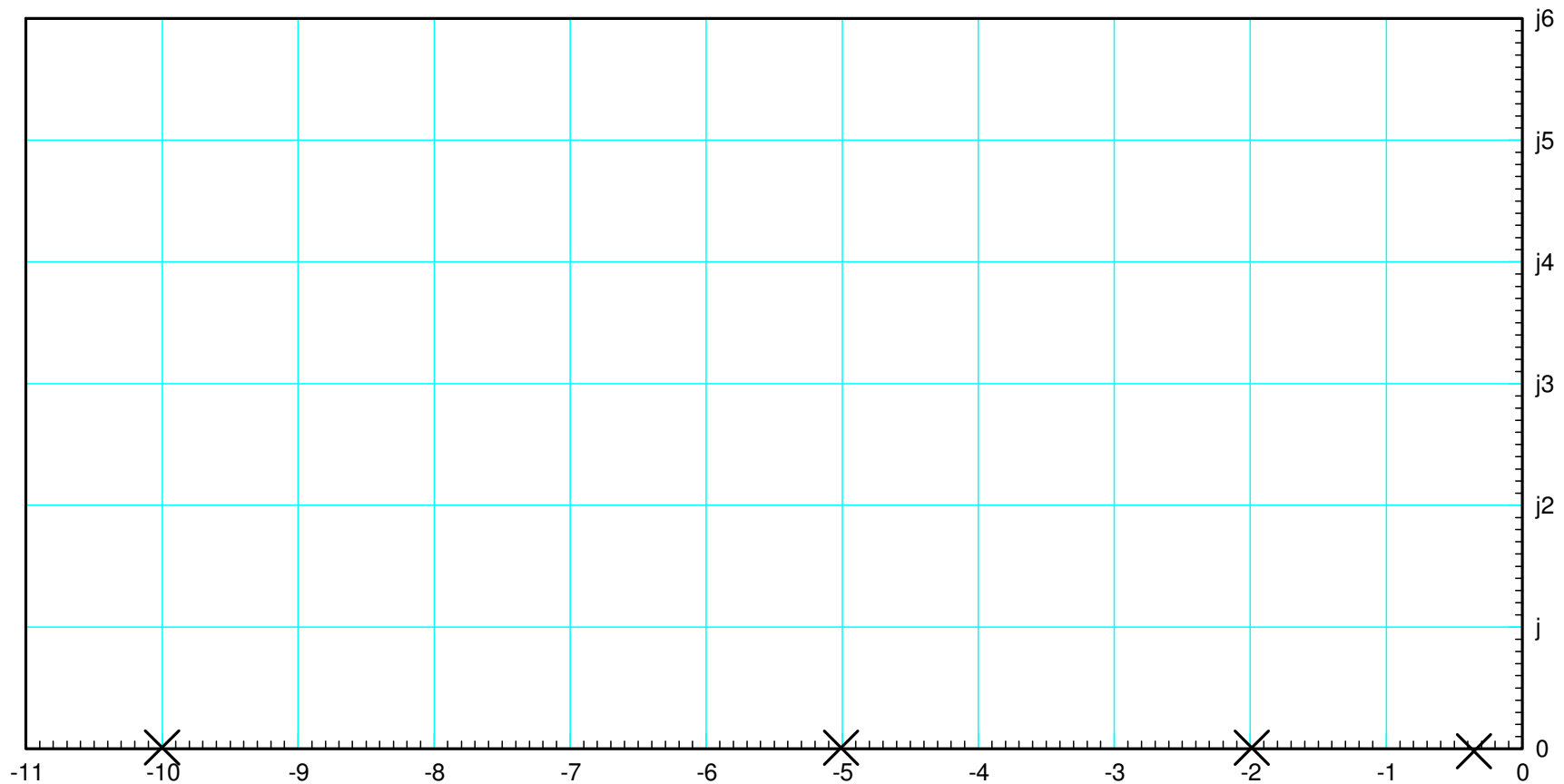


Circuit to implement a 2-stage lead compensator

Handout: Design a lead compensator for the following system

$$G(s) = \left(\frac{200}{(s+0.3)(s+2)(s+5)(s+10)} \right)$$

so that the damping ratio is 0.707 (45 degrees)



Summary:

Lead compensators improve the closed-loop dynamics by pulling the root locus left

Procedure:

- Keep the pole closest to $s=0$
 - Cancel the next slowest stable pole
 - Replace it with a pole 3x to 10x faster
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