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# Using a Transistor as a Switch

**ECE 320 Electronics I**

**Jake Glower - Lecture #12**

Please visit [Bison Academy](#) for corresponding  
lecture notes, homework sets, and solutions

# Using a Transistor as a Switch

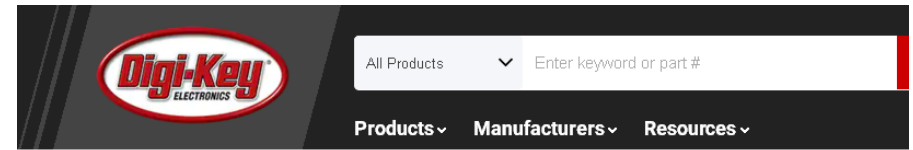
Microcontrollers and function generators (555 timers) have limited outputs

- 0V / 5V up to 20mA

Many loads need more:

- 200mW LED which draws 100mA @ 1.9V
- 8-Ohm speaker which draws 625mA @ 5V.
- Motor which draws 2A at 24V

A transistor can be used as a buffer between the two



Product Index > Integrated Circuits (ICs) > Clock/Timing - Programmable Timers and Oscillators > Harris Corporation CA0555CE

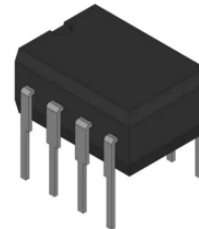




Image shown is a representation only. Exact specifications should be obtained from the product data sheet.

## CA0555CE

Digi-Key Part Number	2156-CA0555CE-ND
Manufacturer	Harris Corporation
Manufacturer Product Number	CA0555CE
Description	TIMER FOR TIMING DELAYS
Detailed Description	555 Type, Timer/Oscillator (Single) IC 8-PDIP
Customer Reference	<input type="text" value="Customer Reference"/>
Datasheet	 <a href="#">Datasheet</a>

## Product Attributes

TYPE	DESCRIPTION	SELECT
Category	<a href="#">Integrated Circuits (ICs)</a> <a href="#">Clock/Timing - Programmable Timers and Oscillators</a>	<input type="radio"/> <input checked="" type="radio"/>
Mfr	Harris Corporation	<input type="checkbox"/>
Series	-	<input type="checkbox"/>
Package	Bulk 	<input type="checkbox"/>
Product Status	Active	<input type="checkbox"/>
Type	555 Type, Timer/Oscillator (Single)	<input type="checkbox"/>
Count	-	<input type="checkbox"/>
Frequency	-	<input type="checkbox"/>

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## On/Off Control with a SPST Switch

Connect your load to power and ground

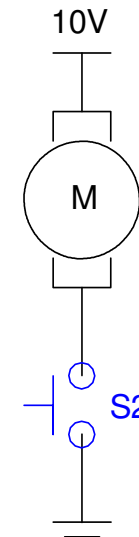
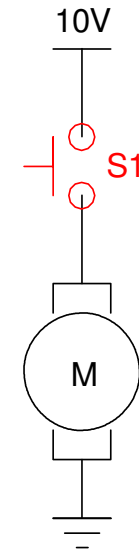
Add a switch,

- At the power supply, or
- At the ground side

Either solution works

Works for any DC load

- Motor (on/off)
- LED (on/off)
- Speaker (push/off)
- etc.



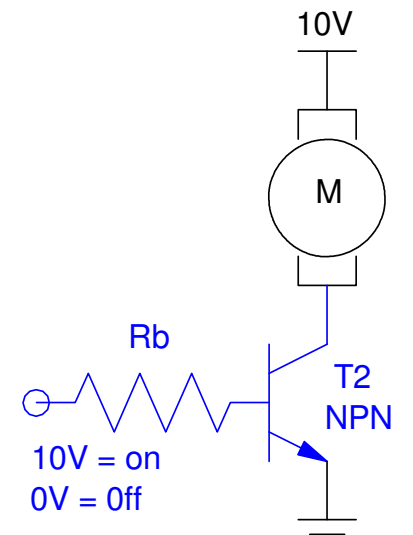
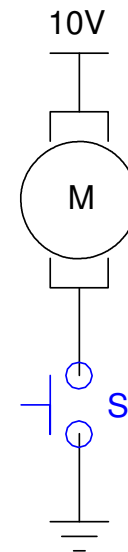
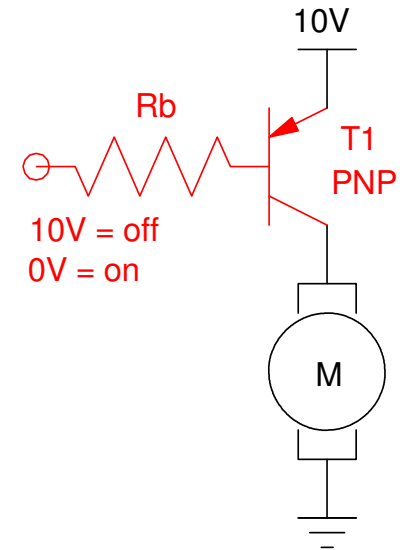
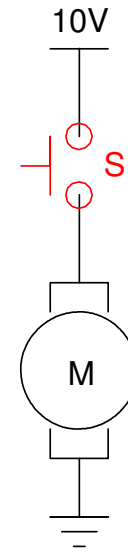
# On/Off Control with a Transistor

Replace the switch with a transistor

- PNP on the high side
- NPN on the low side

Either solution works

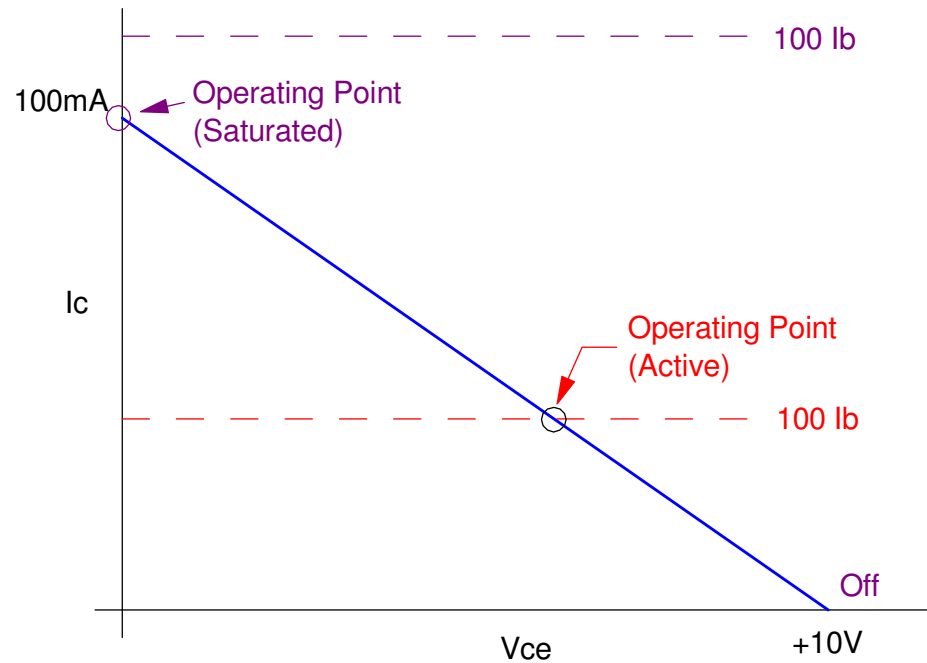
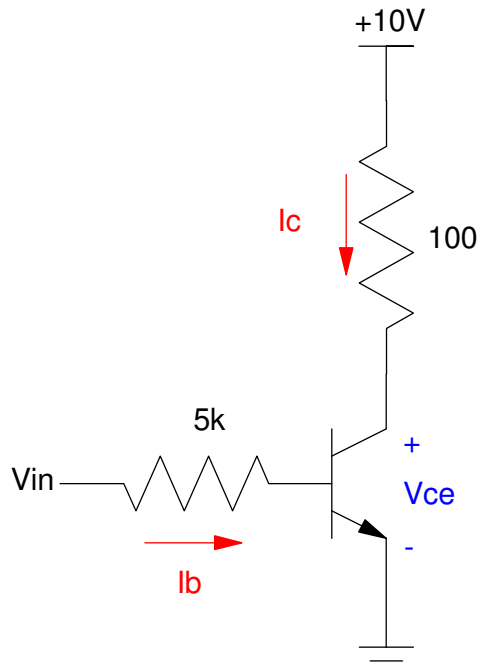
We'll be using the NPN transistor mostly in this class



# NPN Transistor Operation & Load Lines

Transistors are current limiters

- $I_{c(\max)} = \left( \frac{10V - 0.2V}{100} \right) = 98mA$
- $I_c = \min(\beta I_b, I_{c(\max)})$

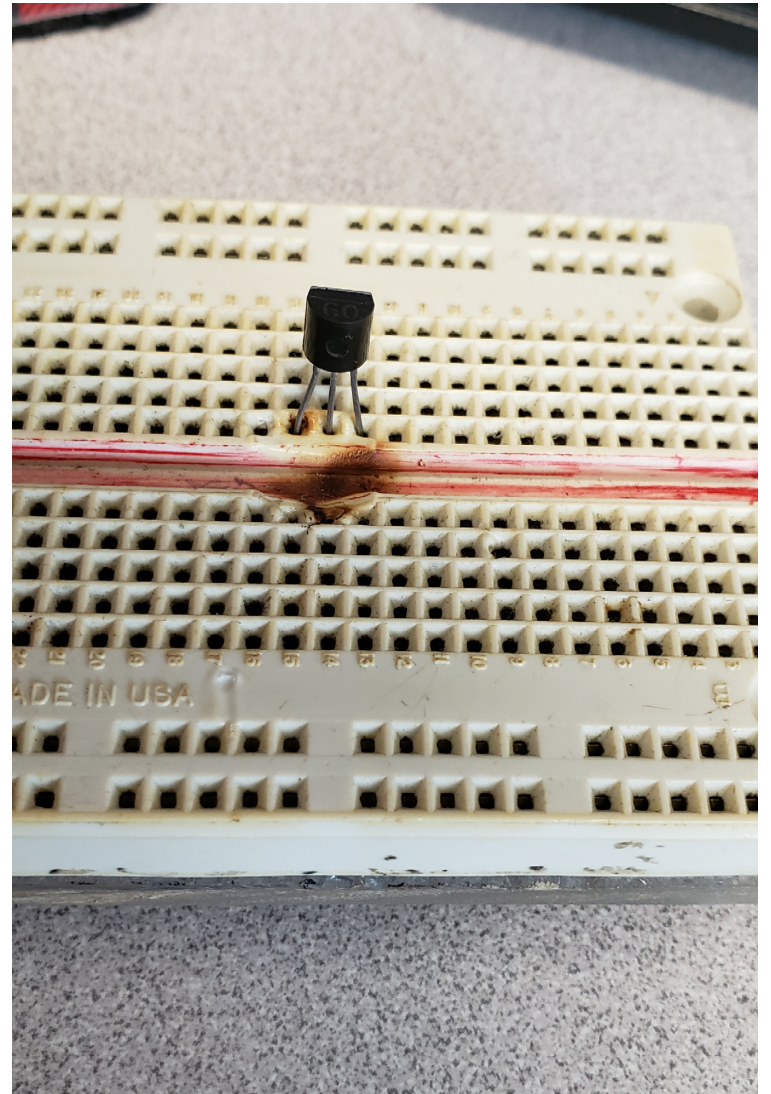


## Goal when Using a Transistor as a Switch:

Operate Either Saturated (on) or Off (off)

- Saturated:
  - Power = 0
  - Good
- Off:
  - Power = 0
  - Good
- Active:
  - Power > 0
  - Bad

If you operate in the active region, the transistor will get hot (and may melt your breadboard)



# Strategy:

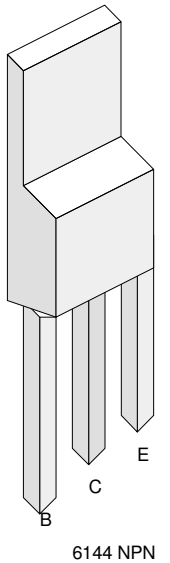
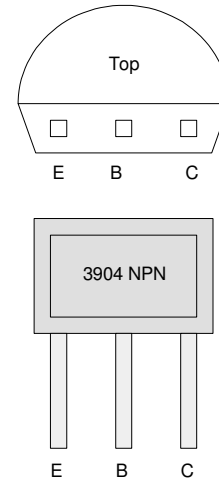
Determine how much current you need

- What is  $I_c$  when the transistor is on?

Pick a transistor which can handle the current

Set  $I_b$  so that

$$\beta I_b > I_c$$



	2N3904 (NPN)	6144 (NPN)
$I_c$ max	200mA	13A (peak) 10A (continuous)
current gain ( $h_{fe}$ = beta)	100 - 300	200 - 560
$V_{be}$ (on)	0.7V	0.7V
$V_{ce}$ (sat)	0.2V	0.18V
Cost	\$0.04 ea	\$0.62 ea

## Example 1: Drive a 200mW LED

Input:

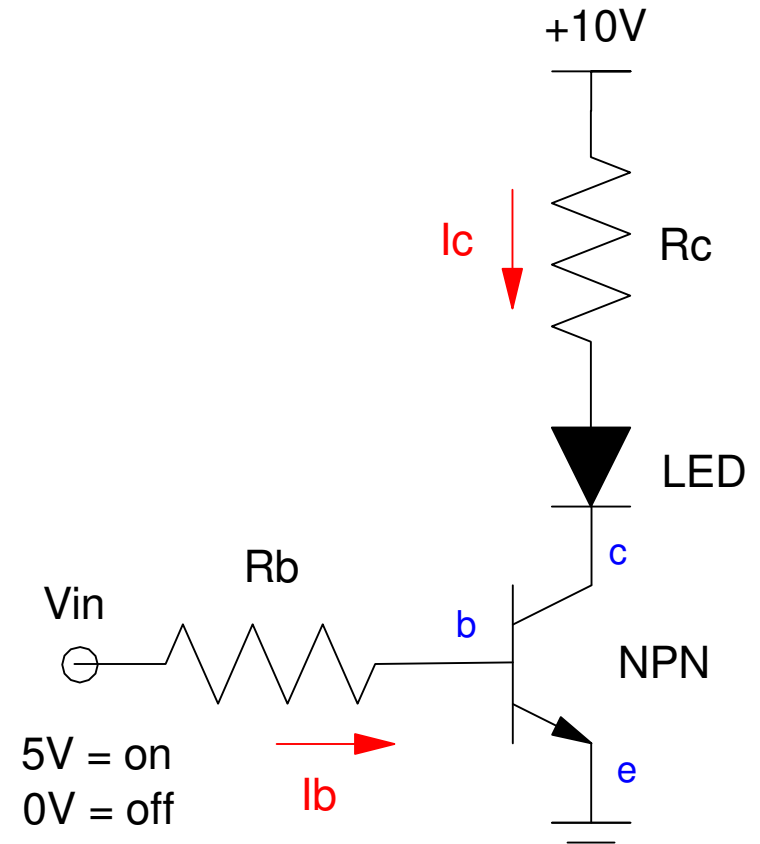
- 0V / 5V capable of 25mA

Output: LED

- $V_f = 1.9V$ , 100mA (max)

Relationship:

- $V_{in} = 0V$ : LED is off
- $V_{in} = 5V$ : LED is on (100mA)





## NPN Solution:

- Pick an NPN transistor that can handle the current.
  - 2N3904 works
- Pick  $R_c$  to set  $I_c = 100\text{mA}$ 
  - The diode drops  $1.9\text{V}$  ( $V_f$ )
  - The transistor drops  $0.2\text{V}$ : ( $V_{ce(\text{sat})}$ )

$$I_c = 100\text{mA} = \left( \frac{10\text{V} - 1.9\text{V} - 0.2\text{V}}{R_c} \right)$$

$$R_c = 79\Omega$$

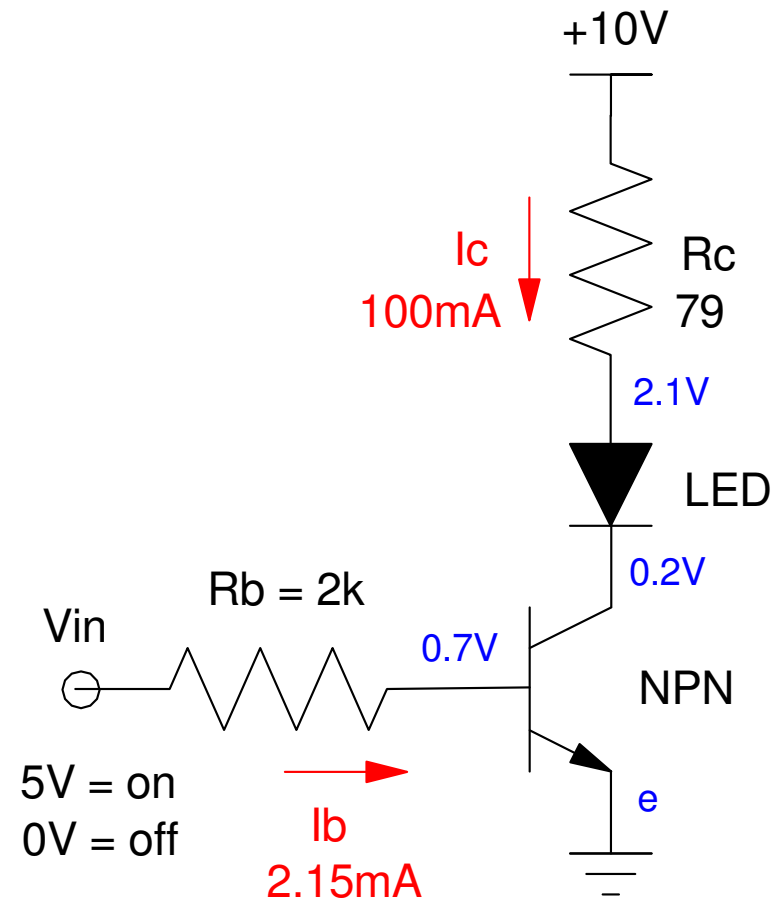
- Pick  $R_b$  so that the transistor saturates

$$\beta I_b > I_c$$

$$I_b > 1\text{mA}$$

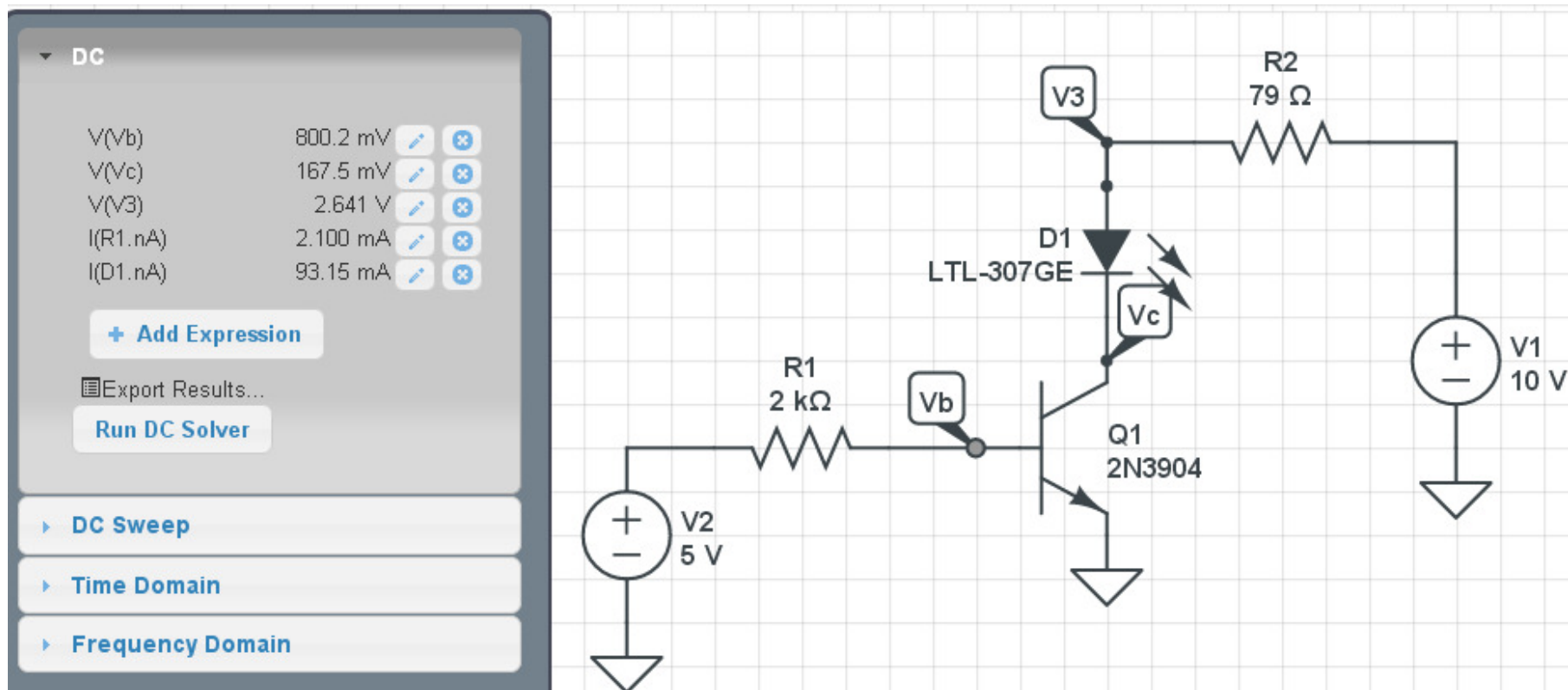
Let  $I_b = 2\text{mA}$

$$R_b = \left( \frac{5\text{V} - 0.7\text{V}}{2\text{mA}} \right) = 2150\Omega \rightarrow 2\text{k}\Omega$$



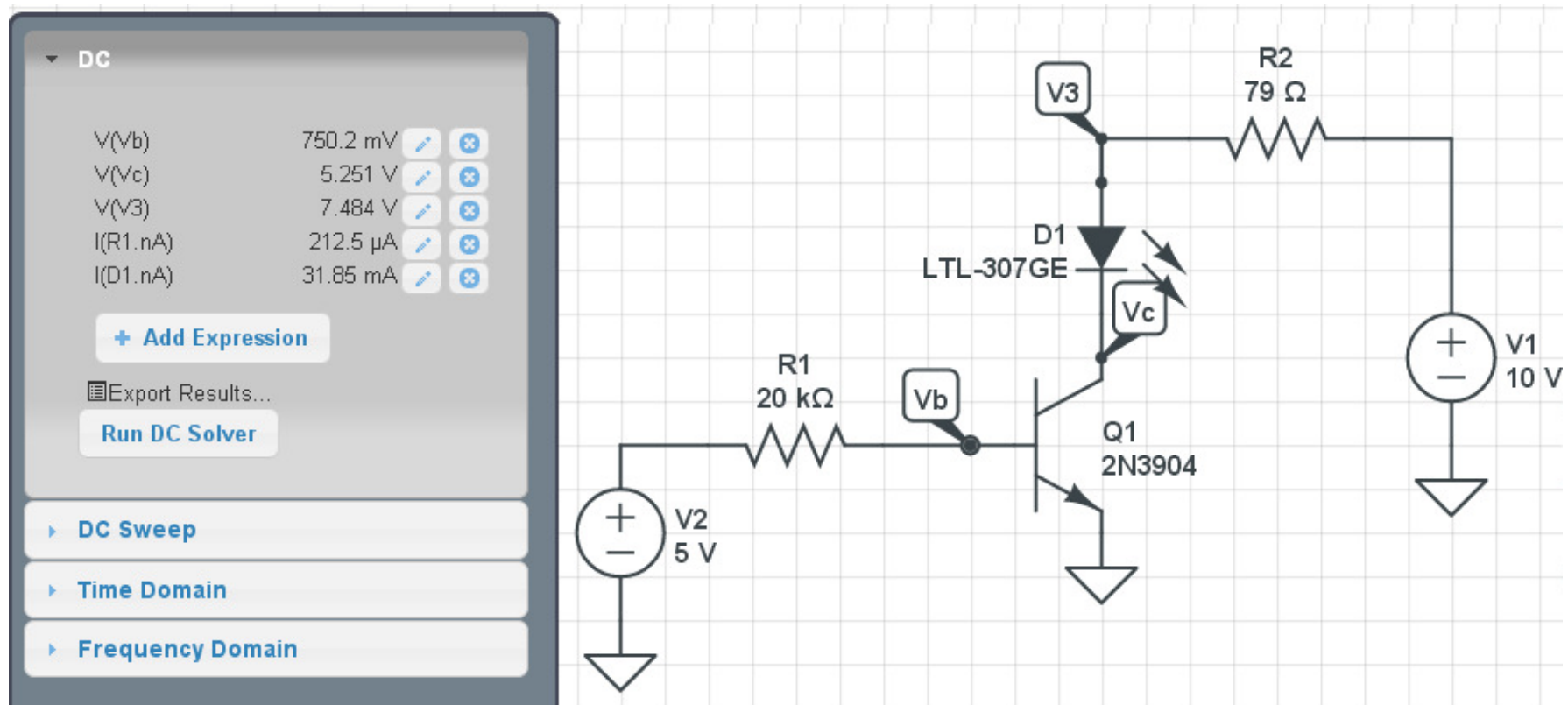
## Checking in CircuitLab

- The voltage across the LED is 2.06V ( had to modify part to get it to approx 1.9V )
- $V_{ce} = 167.5\text{mV}$  ( vs.  $200\text{mV}$  assumed for  $V_{ce}(\text{sat})$  )
- $V_b = 800.2\text{mV}$  ( vs.  $700\text{mV}$  assumed for a saturated silicon diode )
- $I_c = 93.18\text{mA}$  ( vs.  $100\text{mA}$  target )



Note that  $V_{ce} = 167\text{mV}$  tells you that the transistor is saturated:

- $V_{ce} = 200\text{mV}$  is a ballpark guess
  - Close to zero but not quite zero
- If  $V_{ce} > 0.2\text{V}$ , you are operating in the active region
  - You need more base current ( $R_b$  is too large)



# Checking in Hardware

Measure voltages

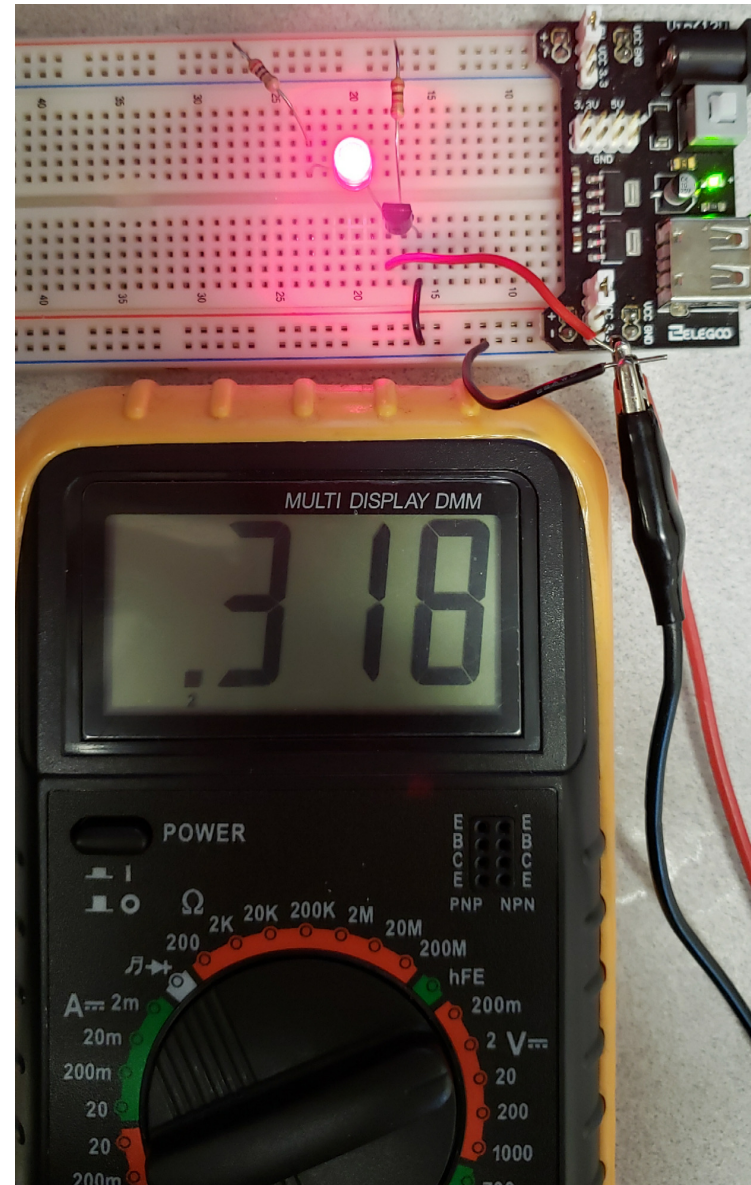
- Calculate currents from  $V = IR$

Results are similar

- Calculated
- Simulated
- Measured

$V_{ce} = 0.2V$  (ish) meaning it's saturated

	$V_{be}$	$V_{ce}$	$I_b$	$I_c$
Calculated	0.7V	0.2V	2.15 mA	100mA
Simulated	0.800 V	0.167V	2.10 mA	93.15 mA
Hardware	0.867V	0.318V	2.07 mA	119.5mA



## Example 2: Drive an 8-Ohm Speaker

Input: 0V / 5V square wave capable of 20mA

Output: 8 Ohm speaker

Relationship:

- When  $V_{in} = 0V$ , the speaker is off (0V and 0mA)
- When  $V_{in} = 5V$ , the speaker is on (4.8V, 600mA)

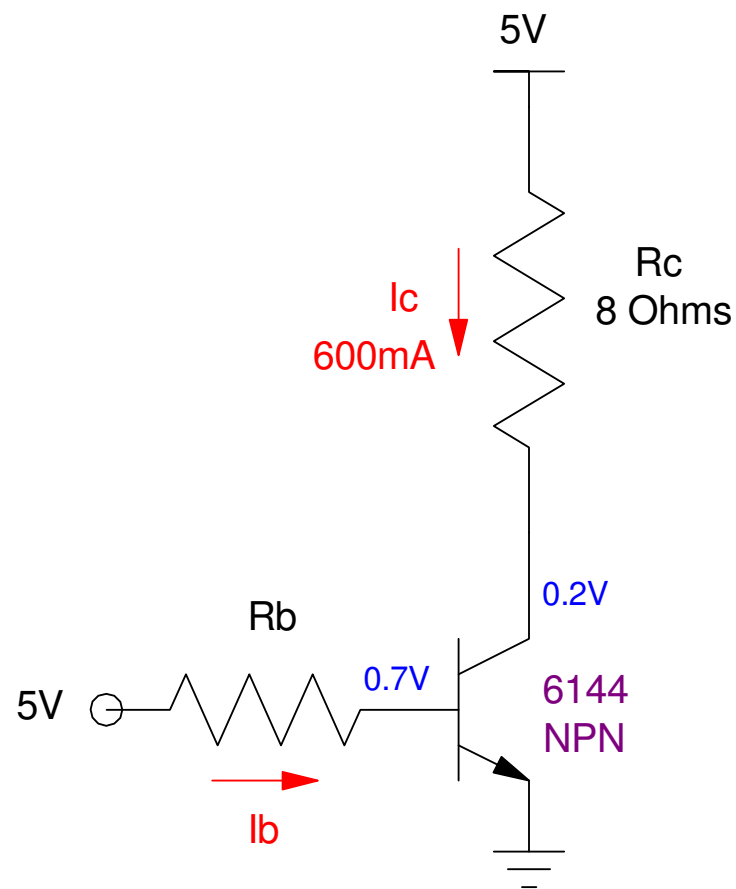
Step 1: Calculate the current you need:

$$I = \left( \frac{5V - 0.2V}{8\Omega} \right) = 600mA$$

Step 2: Pick a transistor that can handle the power

6144 NPN transistor

$$\max(I_c) = 10A$$



Step 3: Determine  $I_b$  ( assume  $I_b < 25\text{mA}$  )

$$\beta I_b > I_c$$

$$25\text{mA} > I_b > \left( \frac{600\text{mA}}{200} \right) = 3\text{mA}$$

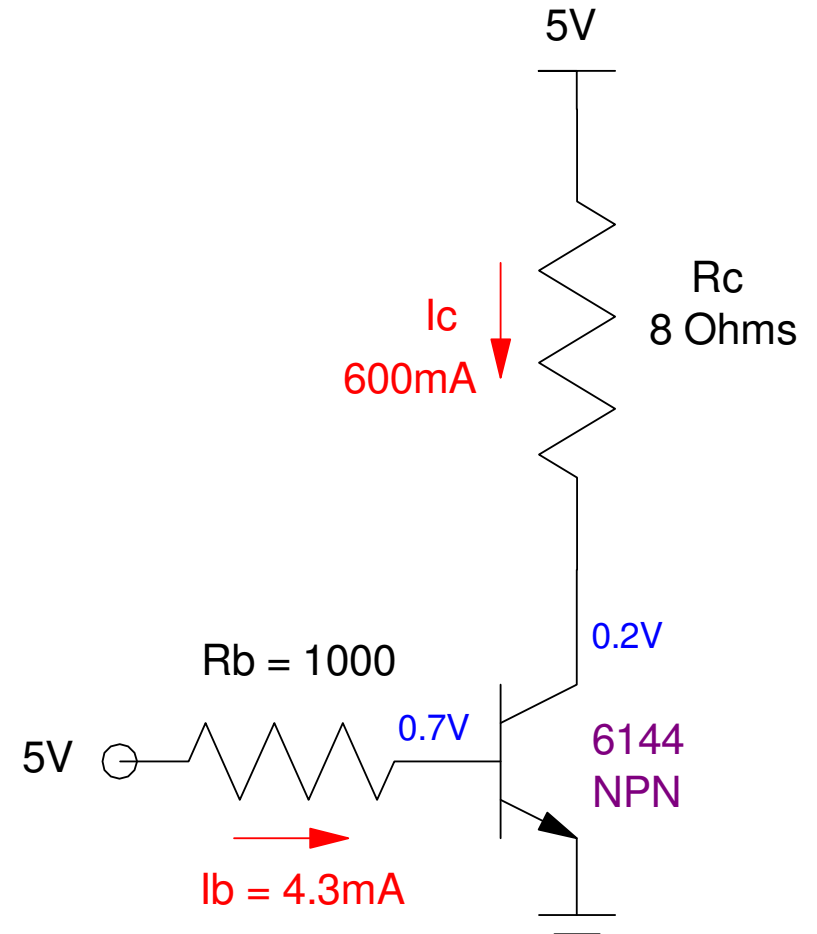
Let  $I_b = 4.3\text{mA}$

Step 4: Determine  $R_b$

$$R_b = \left( \frac{5\text{V} - 0.7\text{V}}{I_b} \right)$$

$$R_b = 1000\Omega$$

Anything in the range of 172 .. 1433 Ohms works



# Checking in CircuitLab

- 6144 NPN transistor isn't an option, so pick one that's close
- Adjust gain to match a 6144 ( $B_F = 300$ )

The screenshot displays the CircuitLab interface with a circuit simulation and a component selection panel.

**Simulation Results:**

V(Vb)	750.2 mV
V(Vc)	5.251 V
V(V3)	7.484 V
I(R1.nA)	212.5 $\mu$ A
I(D1.nA)	31.85 mA

**Buttons:** + Add Expression, Export Results..., Run DC Solver

**DC Sweep**

**Time Domain**

**Frequency Domain**

**Circuit Diagram:** A circuit diagram showing a 5V DC source (V2) connected to a 1 k $\Omega$  resistor (R1). The other end of R1 is connected to the base of an NPN transistor (Q1, TIP41). The base voltage is labeled Vb. The emitter of Q1 is connected to ground. The collector of Q1 is connected to a 5V DC source (V1) through an 8  $\Omega$  resistor (R2). The collector voltage is labeled Vc.

**Component Selection Panel:**

MMBTA06	Datasheet	Buy	name: Q1	Part#: TIP41
MMBTA42	Datasheet	Buy	I <sub>S</sub> : 2.9083e-13 A	B <sub>F</sub> : 300
MPS406	Datasheet	Buy	B <sub>F</sub> : 0.1001	V <sub>A</sub> : 100 V

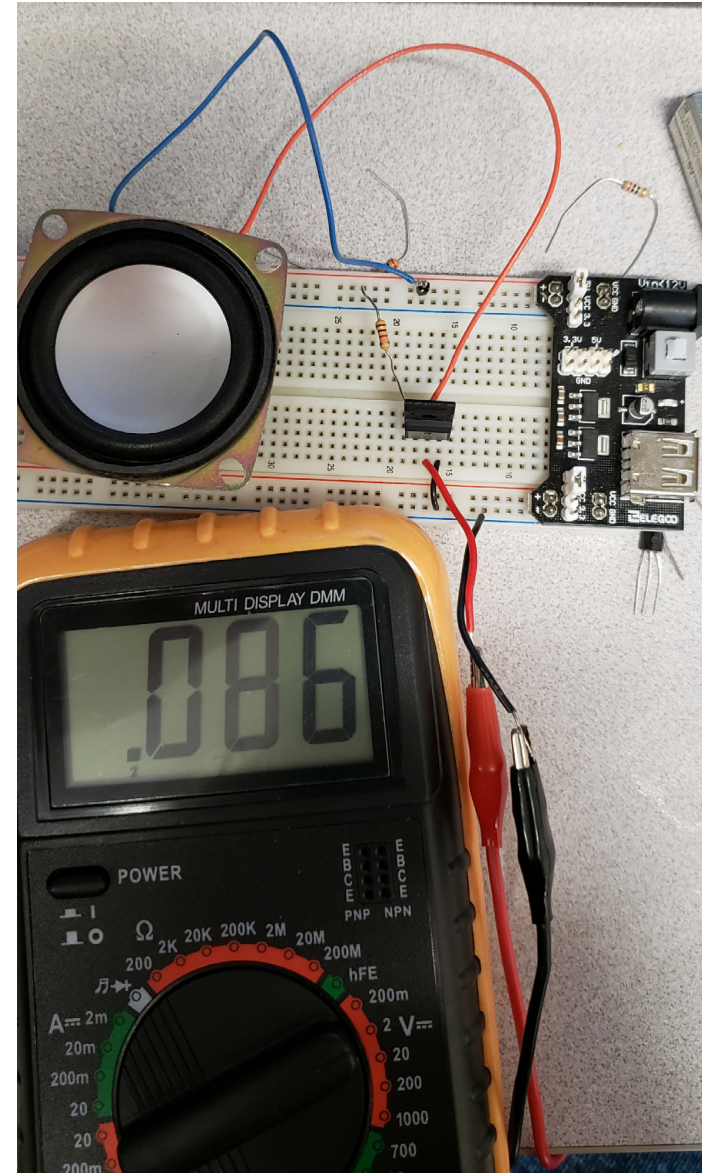


# Checking in Hardware

Measure the voltages

- Calculate currents from  $V = IR$
- $V_{ce} = 0.2V$  (ish) meaning it's saturated

	$V_{be}$	$V_{ce}$	$I_b$	$I_c$
Calculated	0.7V	0.2V	4.30 mA	600.0 mA
Simulated	0.977V	0.088V	4.02 mA	614.0 mA
Hardware	0.670V	0.086V	4.33 mA	614.1mA





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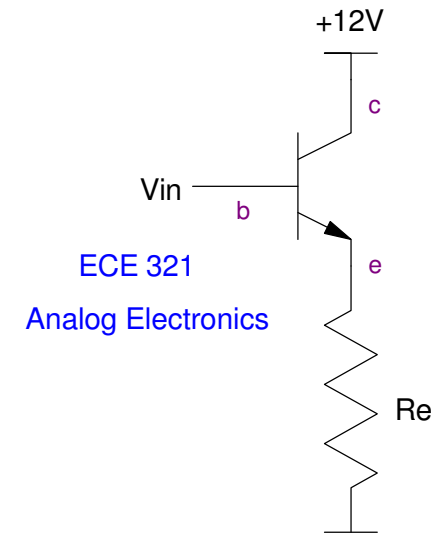
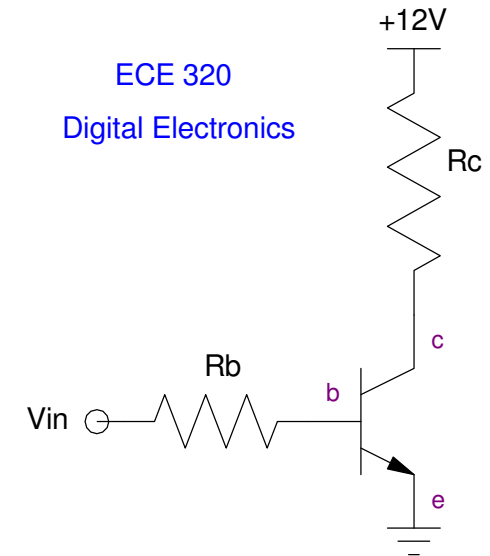
## Note:

In Digital Electronics (ECE 320), the load connects to the collector

- This allows you to operate the transistor in the saturated (on) and off (off) regions
- The transistor acts as a switch

In Analog Electronics (ECE 321), the load connects to the emitter

- This places the transistor in the active region
- The transistor acts as an amplifier

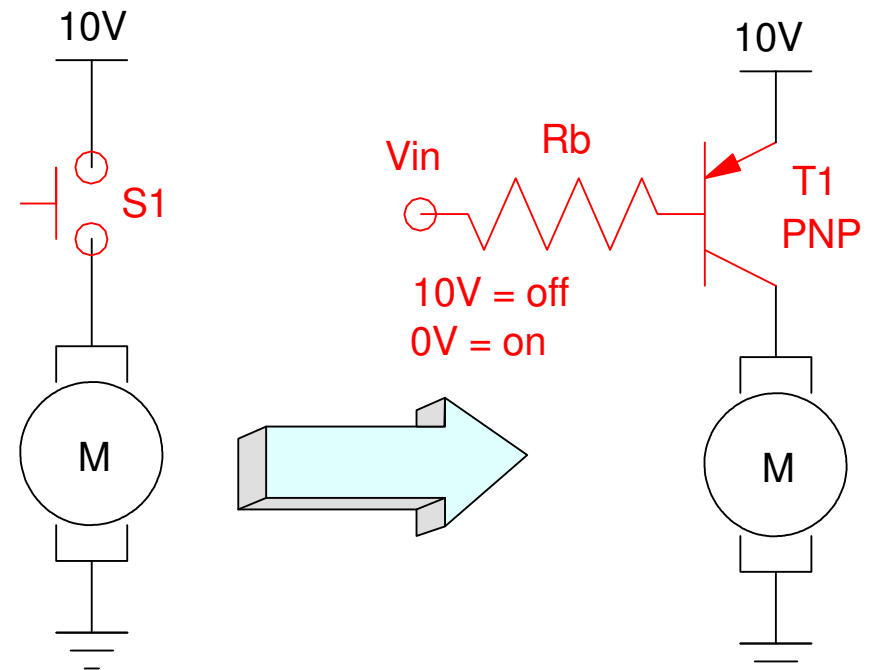


# Electronic Switch with PNP Transistors

- Place the switch on the high side
- Use a PNP transistor as an electronic switch

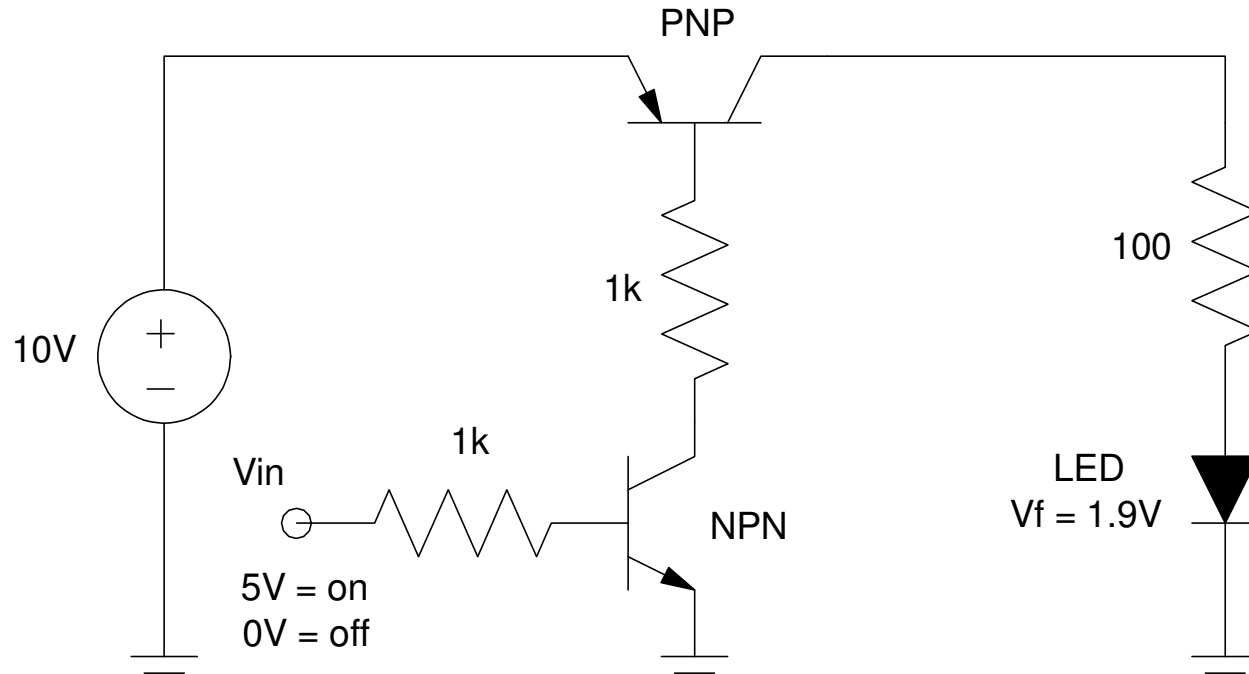
Note: TTL levels don't work

- $V_{in} = 10V$  is off
- $V_{in} = 0V$  is on



## PNP Switch (version 2)

If you want to turn on and off the PNP with a 5V source, add an NPN transistor



# CircuitLab Simulation

▼ DC

V(V0)	5.000 V		
V(V1)	24.39 mV		
V(V2)	9.786 V		
I(R2.nA)	72.55 mA		
I(R3.nA)	9.063 mA		
I(R4.nA)	4.210 mA		

[+ Add Expression](#)

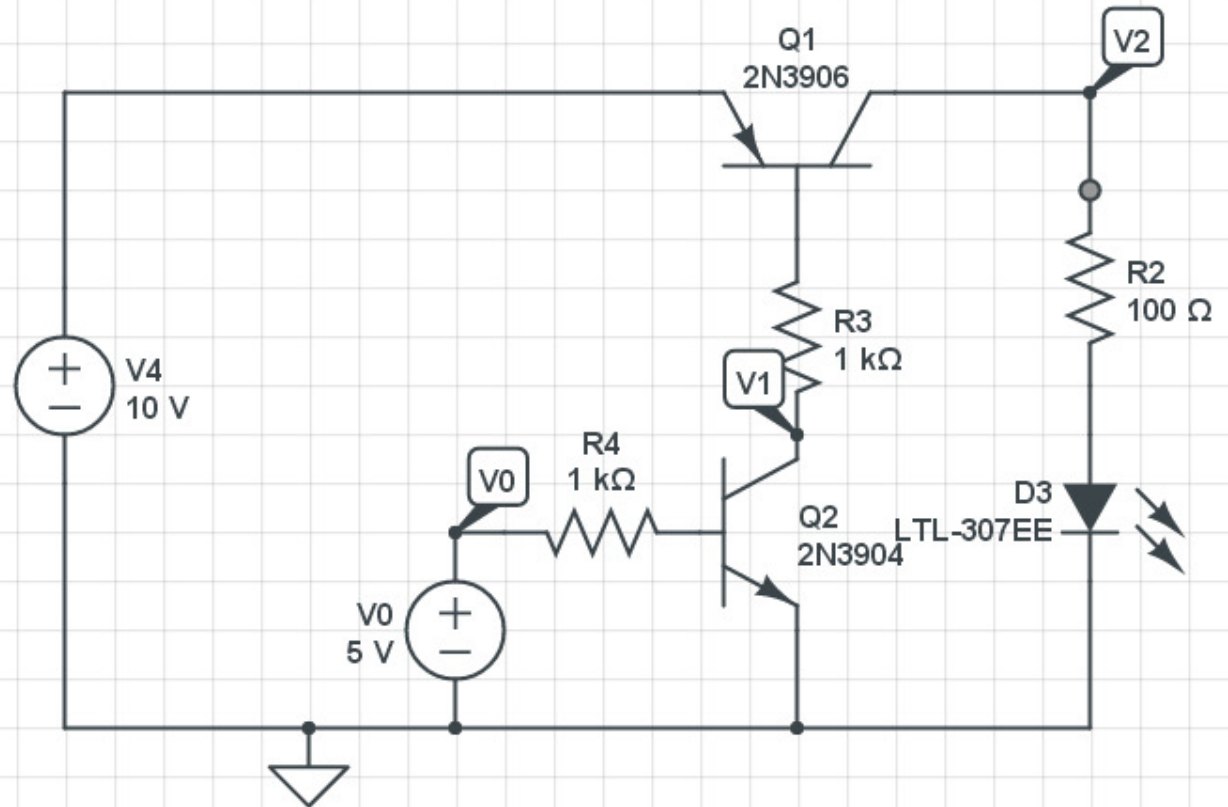
Export Results...

[Run DC Solver](#)

▶ [DC Sweep](#)

▶ [Time Domain](#)

▶ [Frequency Domain](#)



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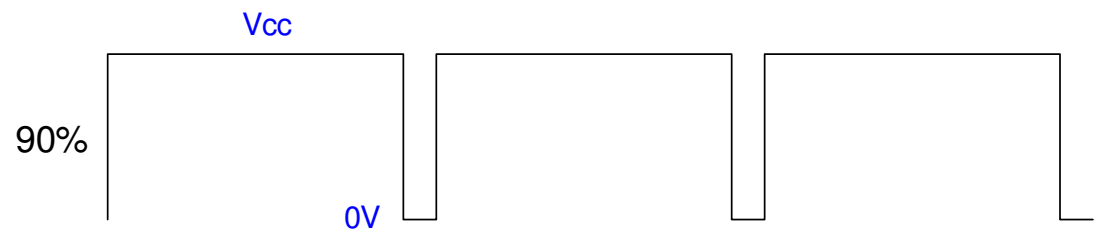
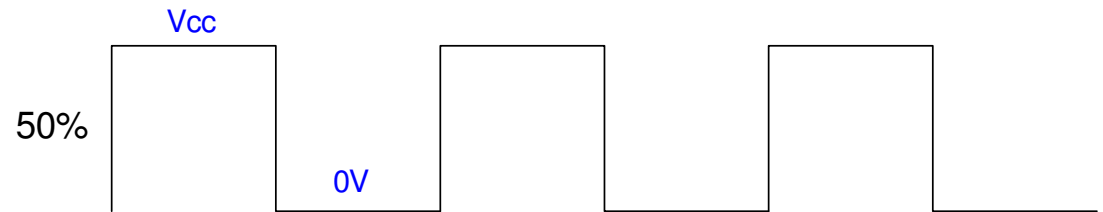
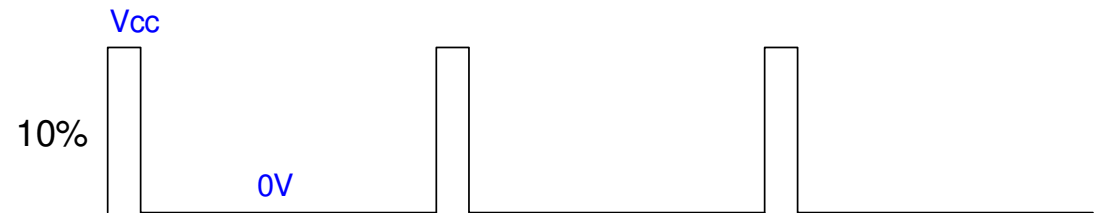
# Pulse Width Modulation (PWM)

You can 'fake' an analog voltage by varying the duty cycle

- $V_{in} = 0V$ : off
- $V_{in} = 5V$ : on
- $V_{in} = X\%$   $X\%$  on

Used in

- Tail-Lights
- RGB scoreboards
- Motor speed control
- etc.



# PWM & LED Brightness Control

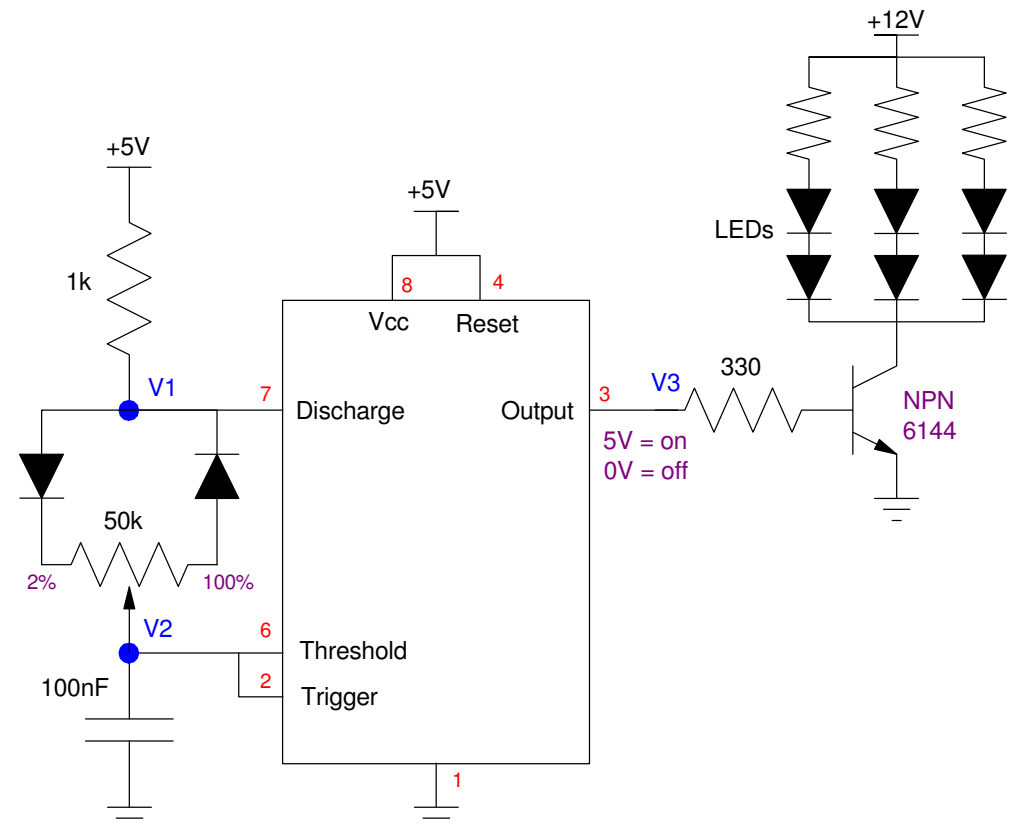
$$V_3 = 5V$$

- $I_b = \left( \frac{5V - 0.7V}{330\Omega} \right) = 13.03mA$
- $I_c = 300I_b = 3.9A$

This power a load up to 3.9A

Brightness is adjustable

- 2% with the pot all the way left
- 50% with the pot in the middle
- 100% with the pot all the way right



# Darlington Pairs (TIP112)

A TIP112 transistor is actually a Darlington pair

- T1 provides gain
- T2 provides high current

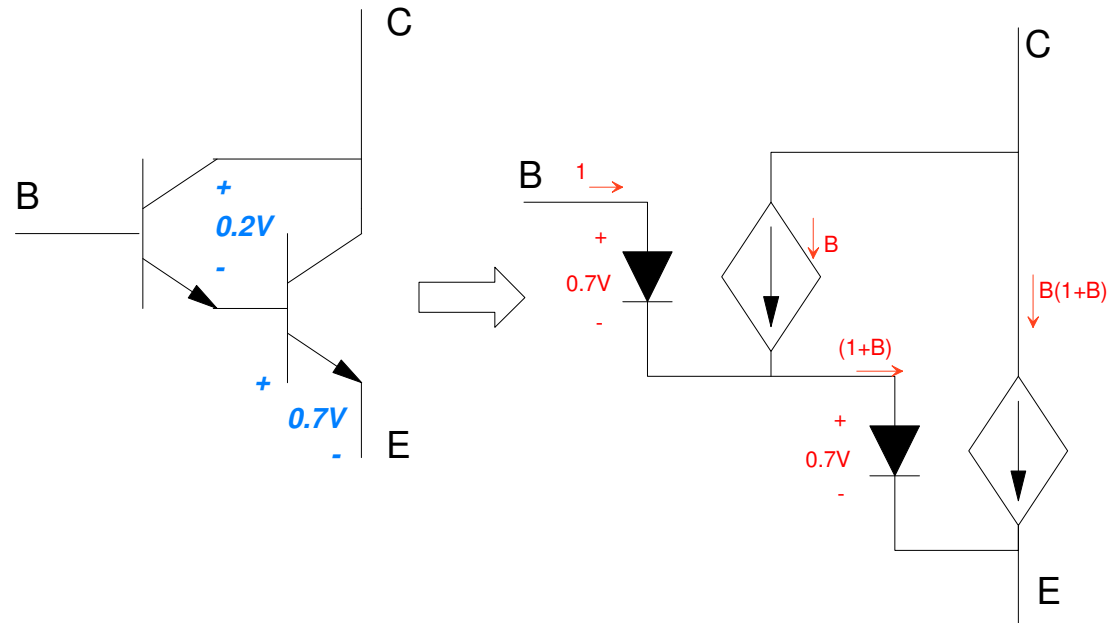
The net result looks like a big hogging transistor

$$V_{be} = 1.4V$$

$$\min V_{ce} = 0.9V$$

$$\beta = (1 + \beta_1)\beta_2 > 1000$$

$$\max(I_c) = 4A$$



## Example 2: Drive an 8-Ohm speaker.

Input: 0V / 5V square wave capable of 20mA

Output: 8 Ohm speaker

Relationship:

- When  $V_{in} = 0V$ , the speaker is off (0V and 0mA)
- When  $V_{in} = 5V$ , the speaker is on (5V, 625mA)

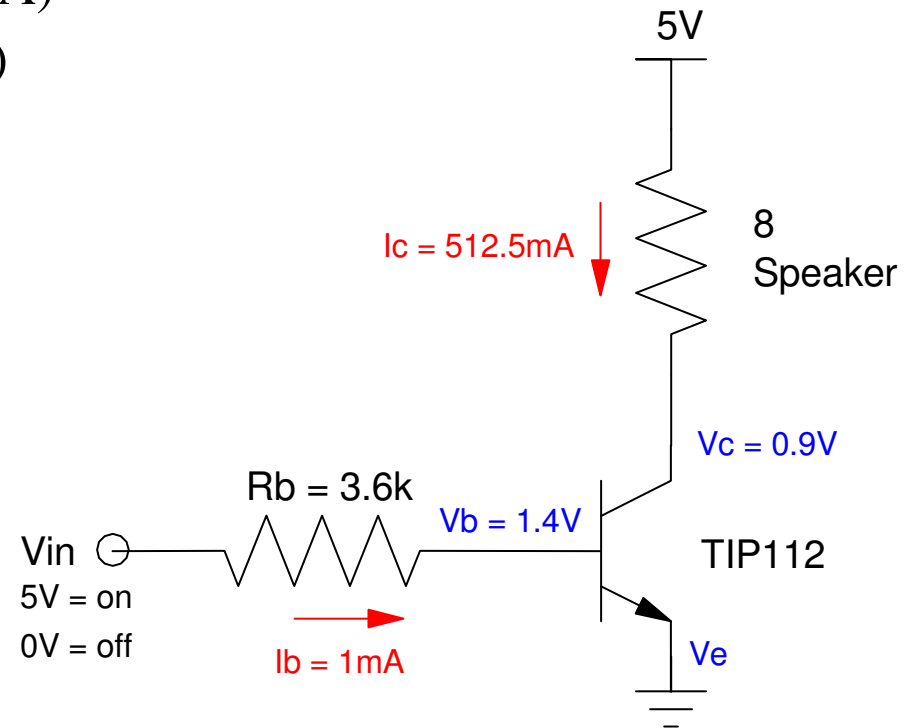
$$I_c = \left( \frac{5V - 0.9V}{8\Omega} \right) = 512.5mA$$

$$\beta I_b > I_c$$

$$I_b > 512.5\mu A$$

Let  $I_b = 1mA$ . Then

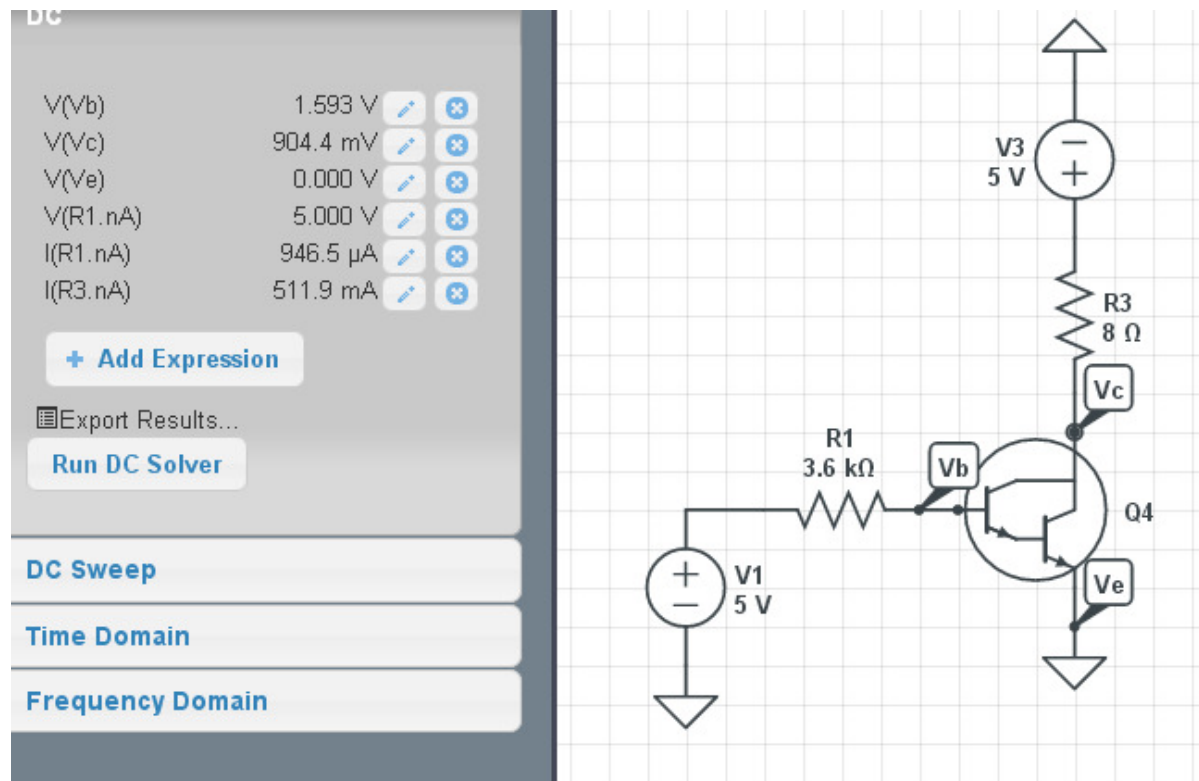
$$R_b = \left( \frac{5V - 1.4V}{1mA} \right) = 3.6k\Omega$$





# CircuitLab Simulation

- $V_b = 1.593\text{V}$       *1.4000V calculated*
- $V_{ce} = 904.4\text{mV}$       *900mV calculated*
- $I_c = 511.9\text{mA}$       *512.5mA calculated*



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# Transistor Switch Demo

How to annoy your friends and neighbors

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# On/Off Control of a DC Motor

Problem:

- A DC motor only draws 30mA @ 24V when operating
- A 2904 transistor is used to turn it on and off, which is capable of 200mA
- The transistor *should* be able to handle the load

The transistor keeps getting fried.

Why?

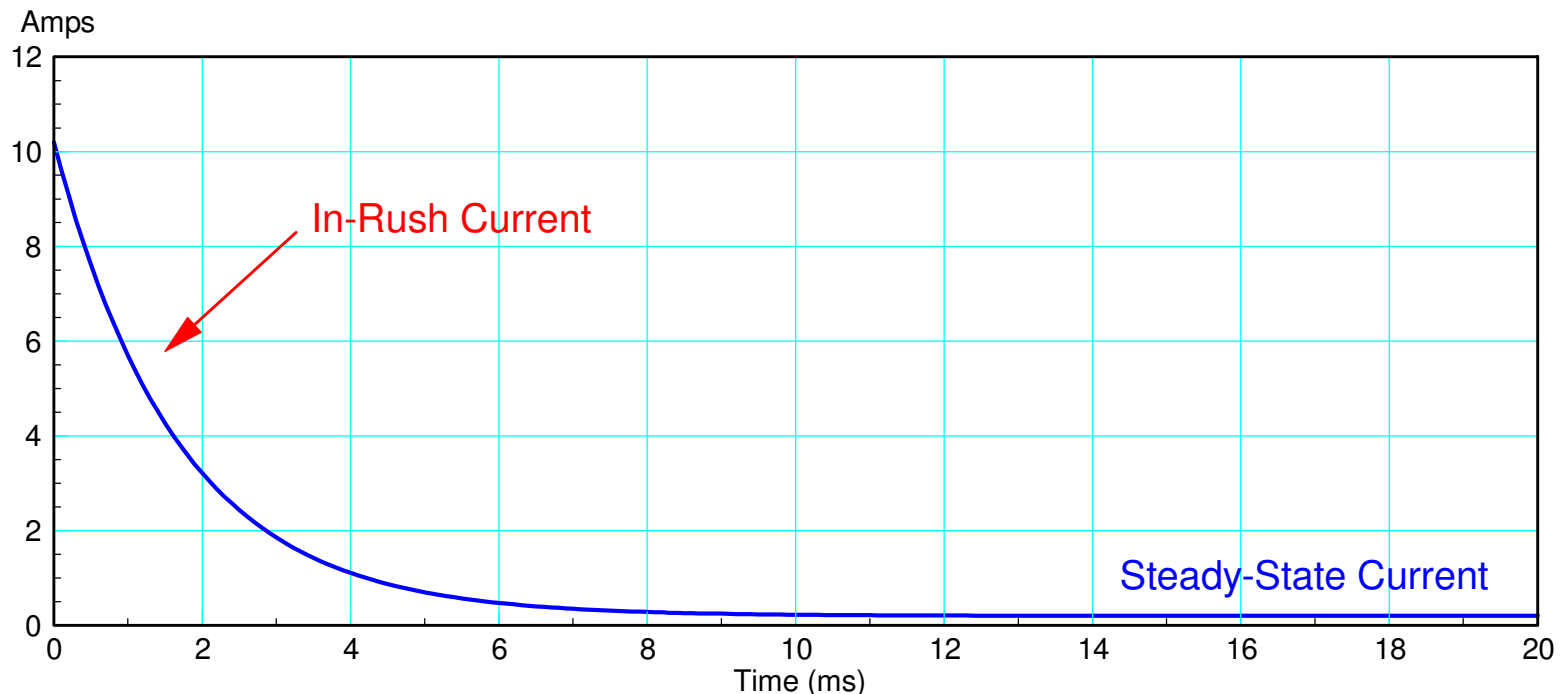


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## Reason #1: In-Rush Current

The start-up current for a DC motor is much larger than the operating current

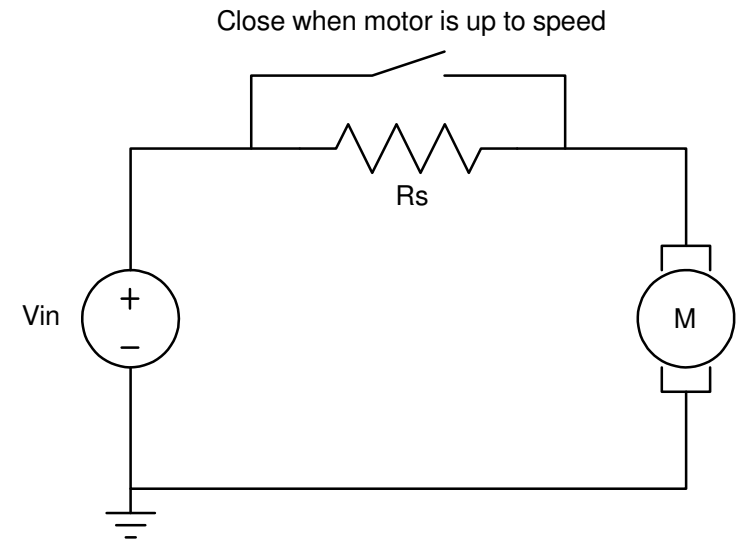
- When starting, a large amount of power is consumed. This power ( $VI$ ) is stored in the motor's kinetic energy. (starting current is large)
- When starting, the back EMF is zero. This results in the current being limited by only the armature resistance. (starting current is large)



# Ways to Limit In-Rush Current

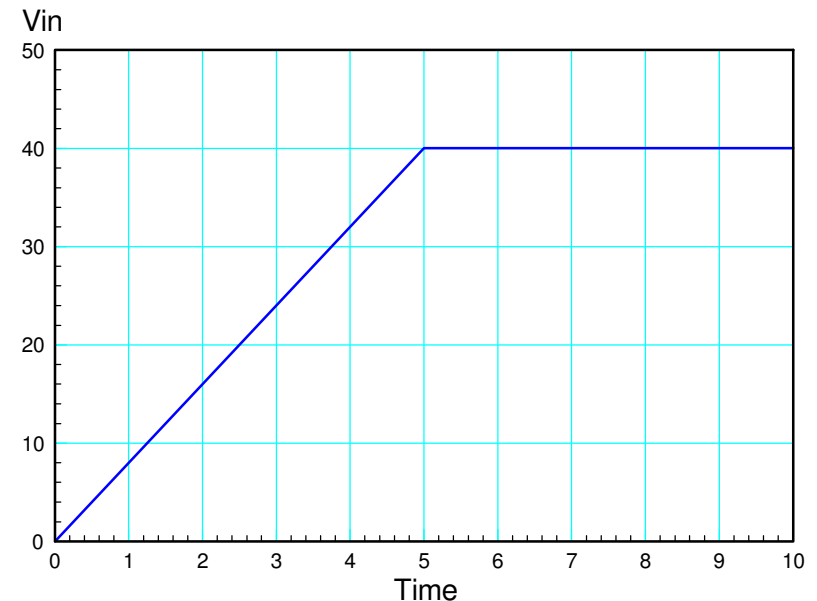
## a) Add a resistor in series during startup

- The in-rush current will burn out the windings in large DC motors (covered in ECE 331)
- By adding a resistor in series, you limit the in-rush current.
- Once the motor gets up to speed, then remove the added resistance.



## b) Ramp up the voltage

- Slowly bring the motor up to speed.



## Reason #2: Motor Inductance

- Motors are inherently inductors.

The voltage across an inductor is

$$V = L \cdot \frac{dI}{dt}$$

Turning off the motor results in  $\frac{dI}{dt} \rightarrow \infty$

- This fries the transistor

This is how spark plugs work

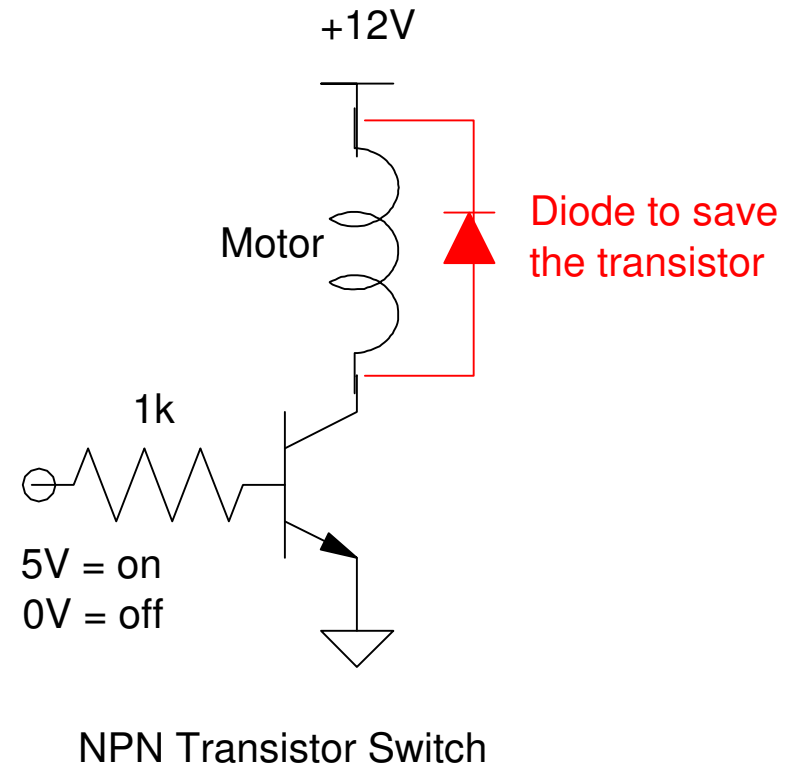
- Pass current through a large inductor (the car's alternator)
- This stores energy in a magnetic field (  $E = \frac{1}{2}LI^2$  )
- When the current goes to zero, the magnetic field collapses
- The energy stored is released by driving the current to infinity



# Fix: Flyback Diodes

## Flyback Diodes

- Provides a path for the current as the magnetic field collapses
- Clips the voltage at  $V_c$  to 12.7V



# Summary

Transistors can be used as an electronic switch

- Pick a transistor that can handle the load
- Place the load on the collector side
- Connect the emitter to ground
- Pick  $R_b$  so that

$$\beta I_b > I_c$$

